

# ILC7083/ILC7084

## 150mA SOT-23 Low Noise CMOS RF-LDO™ Regulator

### Features

- Ultra low 1mV dropout per 1mA load
- 3% output voltage accuracy
- Only 40 $\mu$ V<sub>RMS</sub> noise
- Uses low ESR ceramic output capacitor to minimize noise and output ripple
- Only 100 $\mu$ A ground current at 100mA load
- Ripple rejection up to 85dB at 1kHz, 60dB at 1MHz
- Excellent line and load transient response
- Over current / over temperature protection
- Guaranteed to 150mA output current
- Industry standard five lead SOT-23 package
- Fixed 2.5V, 2.7V, 2.85V, 3.0V, 3.3V, 3.6V, 5V output voltage for ILC7083 and adjustable output voltage for ILC7083/ILC7084
- Metal mask option available for custom voltages between 2.5V and 5.1V

### Applications

- Cellular phones
- Wireless communicators
- PDAs / palmtops / organizers
- Battery powered portable electronics

### Description

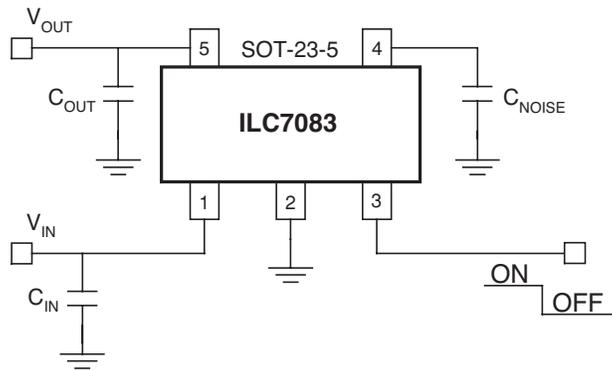
The ILC7083/ILC7084 is a 150mA low dropout (LDO) voltage regulator designed to provide a high performance solution to low power systems. The device offers a typical combination of low dropout and low quiescent current expected of CMOS parts, while uniquely providing the low noise and high ripple rejection characteristics usually only associated with bipolar LDO regulators.

The device has been optimized to meet the needs of modern wireless communications design; Low noise, low dropout, small size, high peak current, high noise immunity.

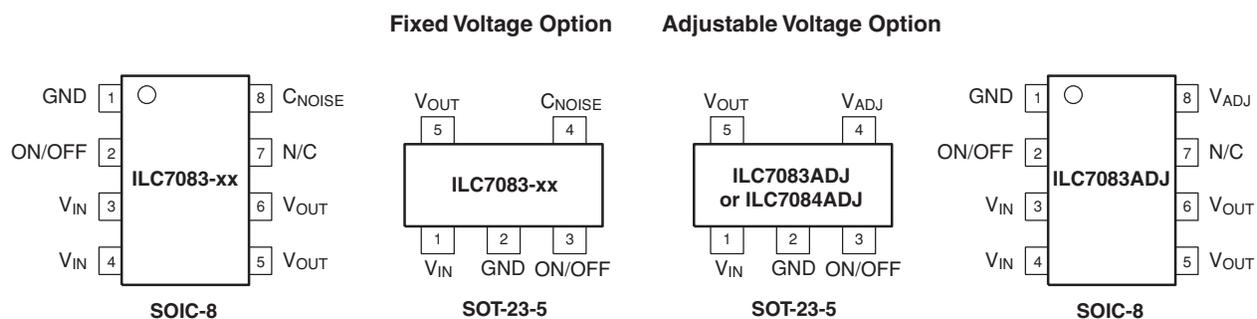
The ILC7083/ILC7084 is designed to make use of low cost ceramic capacitors while outperforming other devices that require tantalum capacitors.

As opposed to ILC7084, the ILC7083 has a built in output capacitor discharge circuit active in shutdown mode. This feature is necessary in applications where the output voltage must decrease quickly to zero volt in shutdown mode.

### Typical Applications



## Pin Assignments



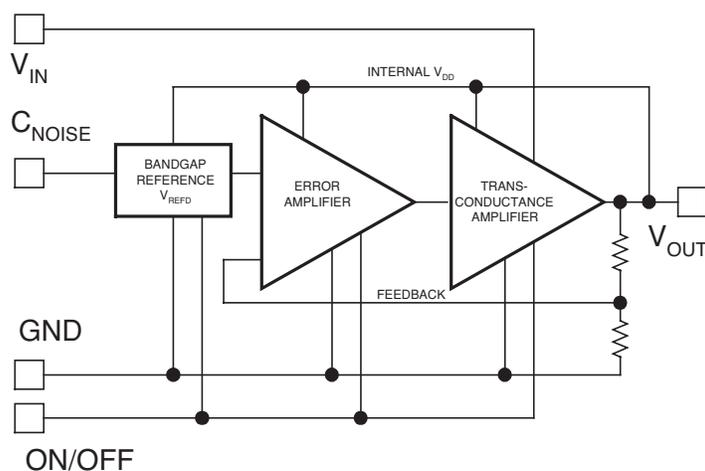
### Pin Description ILC7083-xx (Fixed voltage version)

Pin Number		Pin Name	Pin Description
SOIC-8	SOT-23-5		
3 and 4	1	$V_{IN}$	Connect directly to supply
1	2	GND	Ground pin. Local ground for $C_{NOISE}$ and $C_{OUT}$ .
2	3	ON/OFF	By applying less than 0.6V to this pin the device will be turned off.
8	4	$C_{NOISE}$	Optional noise bypass capacitor may be connected between this pin and GND. Do not connect $C_{NOISE}$ directly to the main power ground plane.
5 and 6	5	$V_{OUT}$	Output Voltage. Connect $C_{OUT}$ between this pin and GND.
7	–	N/C	Not Connected

### Pin Description ILC7083-ADJ (Adjustable voltage version)

Pin Number		Pin Name	Pin Description
SOIC-8	SOT-23-5		
3 and 4	1	$V_{IN}$	Connect directly to supply
1	2	GND	Ground pin. Local ground for $C_{NOISE}$ and $C_{OUT}$ .
2	3	ON/OFF	By applying less than 0.6V to this pin the device will be turned off.
8	4	$V_{ADJ}$	Voltage feedback pin to set the adjustable output voltage. Do not connect a capacitor to this pin.
5 and 6	5	$V_{OUT}$	Output Voltage. Connect $C_{OUT}$ between this pin and GND.
7	–	N/C	Not Connected

## Internal Block Diagram



## Absolute Maximum Ratings

Parameter	Symbol	Ratings	Units
Input Voltage	$V_{IN}$	-0.3 to +13.5	V
On/Off Input Voltage	$V_{ON/OFF}$	-0.3 to $V_{IN}$	
Output Current	$I_{OUT}$	Short circuit protected	mA
Output Voltage	$V_{OUT}$	-0.3 to $V_{IN} + 0.3$	V
Package Power Dissipation (SOT-23-5)	$P_D$	250 (Internally Limited)	mW
Maximum Junction Temp Range	$T_{J(max)}$	-40 to +150	°C
Storage Temperature	$T_{STG}$	-40 to +125	°C
Package Thermal Resistance	$\theta_{JA}$	333	°C/W

## Recommended Operating Conditions

Parameter	Min.	Typ.	Max.	Units
Input Voltage	$V_{OUT} + V_{DO}$	$V_{OUT} + 1$	13	V
Operating Ambient Temperature	-40		+85	°C

## Electrical Characteristics ILC7083/ILC7084

Unless otherwise specified, all limits are at  $T_A=25^\circ\text{C}$ ;  $V_{IN} = V_{OUT(NOM)} + 1\text{V}$ ,  $I_{OUT} = 1\text{mA}$ ,  $C_{OUT} = 1\mu\text{F}$ ,  $V_{ON/OFF} = 2\text{V}$ .  
 Boldface type denotes specifications which apply over the specified operating temperature range.

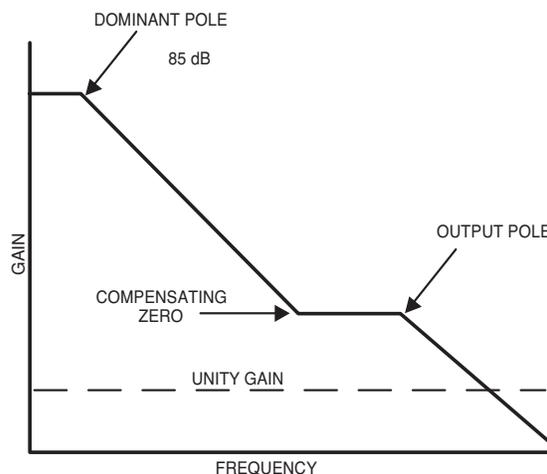
Parameter	Symbol	Conditions	Min	Typ	Max	Units
Input Voltage Range	$V_{IN}$		<b>2</b>		<b>13</b>	V
Output Voltage	$V_{OUT}$	$1\text{mA} \leq I_{OUT} \leq 150\text{mA}$	-3	$V_{OUT(NOM)}$	+3	% $V_{OUT}$ (NOM)
		$1\text{mA} \leq I_{OUT} \leq 150\text{mA}$	<b>-4</b>	$V_{OUT(NOM)}$	<b>+4</b>	
Feedback Voltage (ADJ version)	$V_{ADJ}$		1.217 <b>1.204</b>	1.255	1.292 <b>1.305</b>	V
Line Regulation	$\frac{\Delta V_{OUT}}{(V_{OUT} \cdot \Delta V_{IN})}$	$V_{OUT(NOM)} + 1\text{V} \leq V_{IN} \leq 12\text{V}$		0.007	0.014 <b>0.032</b>	%/V
Dropout Voltage (Note 3)	$V_{IN} - V_{OUT}$	$I_{OUT} = 0\text{mA}$ (Note 4)		0.1	1 <b>2</b>	mV
		$I_{OUT} = 10\text{mA}$		10	25 <b>35</b>	
		$I_{OUT} = 50\text{mA}$		50	75 <b>100</b>	
		$I_{OUT} = 100\text{mA}$		100	150 <b>200</b>	
		$I_{OUT} = 150\text{mA}$		150	225 <b>300</b>	
Ground Pin Current	$I_{GND}$	$I_{OUT} = 0\text{mA}$		95	200 <b>220</b>	$\mu\text{A}$
		$I_{OUT} = 10\text{mA}$		100	220 <b>240</b>	
		$I_{OUT} = 50\text{mA}$		100	220 <b>240</b>	
		$I_{OUT} = 100\text{mA}$		100	240 <b>260</b>	
		$I_{OUT} = 150\text{mA}$		115	260 <b>280</b>	
Shutdown (OFF) Current	$I_{ON/OFF}$	$V_{ON/OFF} = 0\text{V}$		0.1	<b>2</b>	$\mu\text{A}$
ON/OFF Input Voltage	$V_{ON/OFF}$	High = Regulator On Low = Regulator Off	<b>1.5</b>		<b>0.6</b>	
ON/OFF Pin Input Current	$I_{IN(ON/OFF)}$	$V_{ON/OFF} = 0.6\text{V}$ , regulator OFF $V_{ON/OFF} = 2\text{V}$ , regulator ON		0.3 1		$\mu\text{A}$
Peak Output Current (Note 4)	$I_{OUT(peak)}$	$V_{OUT} \geq 0.95V_{OUT(NOM)}$ , tpw = 2ms	400	500		mA
Output Noise Voltage (RMS)	eN	BW = 300Hz to 50kHz, $C_{IN} = 1\mu\text{F}$ $C_{NOISE} = 0.01\mu\text{F}$ , $C_{OUT} = 2.2\mu\text{F}$ , $I_{OUT} = 10\text{mA}$		40		$\mu\text{V}_{RMS}$
Ripple Rejection	$\Delta V_{OUT}/\Delta V_{IN}$	$C_{OUT} = 4.7\mu\text{F}$ , $I_{OUT} = 100\text{mA}$	Freq. = 1kHz		85	dB
			Freq. = 10kHz		70	
			Freq. = 1MHz		60	
Dynamic Line Regulation	$\Delta V_{OUT(line)}$	$V_{IN}$ : $V_{OUT(NOM)} + 1\text{V}$ to $V_{OUT(NOM)} + 2\text{V}$ , tr/ta = 2 $\mu\text{s}$ ; $I_{OUT} = 150\text{mA}$		14		mV
Dynamic Load Regulation	$\Delta V_{OUT(load)}$	$I_{OUT}$ : 1mA to 150mA; tr < 5 $\mu\text{s}$		40		mV
Short Circuit Current	$I_{SC}$	$V_{OUT} = 0\text{V}$		600		mA

### Notes:

- Absolute maximum ratings indicate limits which when exceeded may result in damage to the component. Electrical specifications do not apply when operating the device outside of its rated operating conditions.
- Specified Min/Max limits are production tested or guaranteed through correlation based on statistical control methods. Measurements are taken at constant junction temperature as close to ambient as possible using low duty pulse testing.
- Dropout voltage is defined as the input to output differential voltage at which the output voltage drops 2% below the nominal value measured with an IV differential.
- Guaranteed by design.

## Operation

The ILC7083/ILC7084 LDO design is based on an advanced circuit configuration for which patent protection has been applied. Typically it is very difficult to drive a capacitive output with an amplifier. The output capacitance produces a pole in the feedback path, which upsets the carefully tailored dominant pole of the internal amplifier. Traditionally the pole of the output capacitor has been “eliminated” by reducing the output impedance of the regulator such that the pole of the output capacitor is moved well beyond the gain bandwidth product of the regulator. In practice, this is difficult to do and still maintain high frequency operation. Typically the output impedance of the regulator is not simply resistive, such that the reactive output impedance interacts with the reactive impedance of the load resistance and capacitance. In addition, it is necessary to place the dominant pole of the circuit at a sufficiently low frequency such that the gain of the regulator has fallen below unity before any of the complex interactions between the output and the load occur. The ILC7083/ILC7084 does not try to eliminate the output pole, but incorporates it into the stability scheme. The load and output capacitor forms a pole, which rolls off the gain of the regulator below unity. In order to do this the output impedance of the regulator must be high, looking like a current source. The output stage of the regulator becomes a transconductance amplifier, which converts a voltage to a current with a substantial output impedance. The circuit which drives the transconductance amplifier is the error amplifier, which compares the regulator output to the band gap reference and produces an error voltage as the input to the transconductance amplifier. The error amplifier has a dominant pole at low frequency and a “zero” which cancels out the effects of the pole. The zero allows the regulator to have gain out to the frequency where the output pole continues to reduce the gain to unity. The configuration of the poles and zero are shown in Figure 1. Instead of powering the critical circuits from the unregulated input voltage, the CMOS RF LDO powers the internal circuits such as the bandgap, the error amplifier and most of the transconductance amplifier from the boot strapped regulated output voltage of the regulator. This technique offers extremely high ripple rejection and excellent line transient response.



**Figure 1. ILC7083/ILC7084 RF LDO Frequency Response**

A block diagram of the regulator circuit used in the ILC7083 is shown in Figure 2, which shows the input-to-output isolation and the cascaded sequence of amplifiers that implement the pole-zero scheme outlined above.

The ILC7083/ILC7084 is designed in a CMOS process with some minor additions, which allow the circuit to be used at input voltages up to 13V. The resulting circuit exceeds the frequency response of traditional bipolar circuits. The ILC7083/ILC7084 is very tolerant of output load conditions with the inclusion of both short circuit and thermal overload protection. The device has a very low dropout voltage, typically a linear response of 1mV per milliamp of load current, and none of the quasi-saturation characteristics of a bipolar output devices. All the good features of the frequency response and regulation are valid right to the point where the regulator goes out of regulation in a 4 millivolt transition region. Because there is no base drive, the regulator is capable of providing high current surges while remaining in regulation. This is shown in the high peak current of 500mA which allows for the ILC7083/ILC7084 to be used in systems that require short burst mode operation.

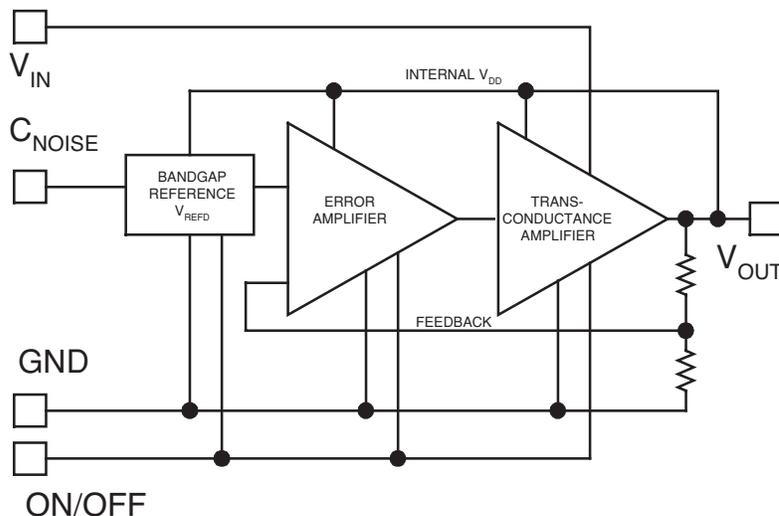


Figure 2. ILC7083 RF LDO Regulator Block Diagram

### Shutdown (ON/OFF) Operation

The ILC7083/ILC7084 output can be turned off by applying 0.6V or less to the device’s ON/OFF pin (pin 3). In shutdown mode, the ILC7083/ILC7084 draws less than 1µA quiescent current. In shutdown mode the output of ILC7083 is coupled to a low impedance circuit which discharges the output capacitor. The output of the ILC7083/ILC7084 is enabled by applying 1.5V to 13V at the ON/OFF pin. In applications where the ILC7083 output will always remain enabled, the ON/OFF pin may be connected to V\_IN (pin 1). The ILC7083/ILC7084’s shutdown circuitry includes hysteresis, as such the device will operate properly even if a slow moving signal is applied to the ON/OFF pin.

### Short Circuit Protection

The ILC7083/ILC7084 output can withstand momentary short circuit to ground. Moreover, the regulator can deliver very high output peak current due to its 1A instantaneous short circuit current capability.

### Thermal Protection

The ILC7083/ILC7084 also includes a thermal protection circuit which shuts down the regulator when die temperature exceeds 150°C due to overheating. In thermal shutdown, once the die temperature cools to below 140°C, the regulator is enabled. If the die temperature is excessive due to high package power dissipation, the regulator’s thermal circuit will continue to pulse the regulator on and off. This is called thermal cycling.

Excessively high die temperature may occur due to high differential voltage across the regulator or high load current or high ambient temperature or a combination of all three. Thermal protection protects the regulator from such fault conditions and is a necessary requirement in today’s designs.

In normal operation, the die temperature should be limited to under 150°C.

### Adjustable Output Voltage

Figure 5 shows how an adjustable output voltage can be easily achieved using ILC7083/ILC7084-ADJ. The output voltage, V\_OUT is given by the following equation:

$$V_{OUT} = 1.255V \times (R1/R2 + 1)$$

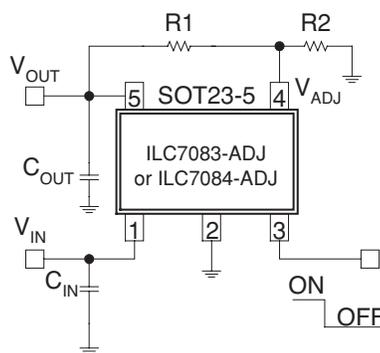


Figure 3. Application Circuit for Adjustable Output Voltage

For best results, a resistor value of 470kΩ or less may be used for R2. The output voltage can be programmed from 2.5V to 12V.

**Note that an external capacitor should not be connected to the adjustable feedback pin (pin 4). Connecting an external capacitor to pin 4 may cause regulator instability and lead to oscillations.**

## Maximum Output Current

The maximum output current available from the ILC7083/ILC7084 is limited by the maximum package power dissipation as well as the device's internal current limit. For a given ambient temperature,  $T_A$ , the maximum package power dissipation is given by:

$$P_{D(MAX)} = (T_{J(MAX)} - T_A) / \theta_{JA}$$

where  $T_{J(MAX)} = 150^\circ\text{C}$  is the maximum junction temperature and  $\theta_{JA} = 333^\circ\text{C/W}$  is the package thermal resistance. For example at  $T_A = 85^\circ\text{C}$  ambient temperature, the maximum package power dissipation is;

$$P_{D(MAX)} = 195\text{mW.}$$

The maximum output current can be calculated from the following equation:

$$I_{OUT(MAX)} < P_{D(MAX)} / (V_{IN} - V_{OUT})$$

For example at  $V_{IN} = 6\text{V}$ ,  $V_{OUT} = 5\text{V}$  and  $T_A = 85^\circ\text{C}$ , the maximum output current is  $I_{OUT(MAX)} < 195\text{mA}$ . At higher output current, the die temperature will rise and cause the thermal protection circuit to be enabled.

## Application Hints

Figure 4 shows the typical application circuit for the ILC7083.

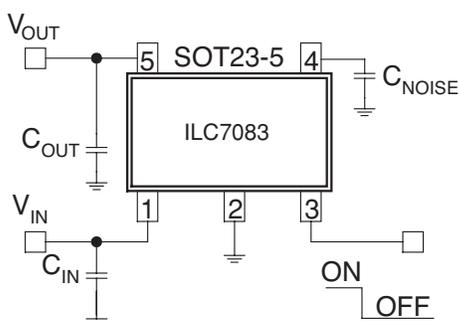


Figure 4. Basic Application Circuit for Fixed Output Voltage Versions

## Input Capacitor

An input capacitor  $C_{IN}$  of value  $1\mu\text{F}$  or larger should be connected from  $V_{IN}$  to the main ground plane. This will help to filter supply noise from entering the LDO. The input capacitor should be connected as close to the LDO regulator input pin as is practical. Using a high-value input capacitor will offer superior line transient response as well as better power supply ripple rejection. A ceramic or tantalum capacitor may be used at the input of the LDO regulator.

Note that there is a parasitic diode from the LDO regulator output to the input. If the input voltage swings below the regulator's output voltage by a couple of hundred millivolts then

the regulator may be damaged. This condition must be avoided. In many applications a large value input capacitor,  $C_{IN}$ , will hold  $V_{IN}$  higher than  $V_{OUT}$  and decay slower than  $V_{OUT}$  when the LDO is powered off.

## Output Capacitor Selection

Fairchild strongly recommends the use of low ESR (equivalent series resistance) ceramic capacitors for  $C_{OUT}$  and  $C_{NOISE}$ . The ILC7083/ILC7084 is stable with low ESR capacitor (as low as zero  $\Omega$ ). The value of the output capacitor should be  $1\mu\text{F}$  or higher. Either ceramic chip or a tantalum capacitor may be used at the output.

Use of ceramic chip capacitors offer significant advantages over tantalum capacitors. A ceramic capacitor is typically cheaper than a tantalum capacitor, it usually has a smaller footprint, lower height, and lighter weight than a tantalum capacitor. Furthermore, unlike tantalum capacitors which are polarized and can be damaged if connected incorrectly, ceramic capacitors are non-polarized. Low value ceramic chip capacitors with X5R or X7R dielectric are available in the  $100\text{pF}$  to  $4.7\mu\text{F}$  range. Beware of using ceramic capacitors with Y5V dielectric since their ESR increases significantly at cold temperatures. Please see the table of recommended ceramic capacitors for use at the output of ILC7083/ILC7084.

Note: If a tantalum output capacitor is used then for stable operation. We recommend a low ESR tantalum capacitor with maximum rated ESR at or below  $0.4\Omega$ . Low ESR tantalum capacitors, such as the TPS series from AVX Corporation ([www.avxcorp.com](http://www.avxcorp.com)) or the T495 series from Kemet ([www.kemet.com](http://www.kemet.com)) may be used.

In applications where a high output surge current can be expected, use a high value but low ESR output capacitor for superior load transient response. The ILC7083/ILC7084 is stable with no load.

## Noise Bypass Capacitor

In low noise applications, the self noise of the ILC7083 can be decreased further by connecting a capacitor from the noise bypass pin (pin 4) to ground (pin 2). The noise bypass pin is a high impedance node as such care should be taken in printed circuit board layout to avoid noise pick-up from external sources. Moreover, the noise bypass capacitor should have low leakage.

Noise bypass capacitors with a value as low as  $470\text{pF}$  may be used. However, for optimum performance, use a  $0.01\mu\text{F}$  or larger, ceramic chip capacitor. Note that the turn on and turn off response of the ILC7083 is inversely proportional to the value of the noise bypass capacitor. For fast turn on and turn off, use a small value noise bypass capacitor. In applications where exceptionally low output noise is not required, consider omitting the noise bypass capacitor altogether.

### The Effects of ESR (Equivalent Series Resistance)

The ESR of a capacitor is a measure of the resistance due to the leads and the internal connections of the component. Typically measured in mΩ (milli-ohms) it can increase to ohms in some cases.

Wherever there is a combination of resistance and current, voltages will be present. The control functions of LDOs use two voltages in order to maintain the output precisely;  $V_{OUT}$  and  $V_{REF}$ .

With reference to the block diagram in Figure 4,  $V_{OUT}$  is fed back to the error amplifier and is used as the supply voltage for the internal components of the ILC7083/ILC7084. So any change in  $V_{OUT}$  will cause the error amplifier to try to compensate to maintain  $V_{OUT}$  at the set level and noise on  $V_{OUT}$  will be reflected into the supply of each internal component of the ILC7083/ILC7084. So any change in  $V_{OUT}$  will cause the error amplifier to try to compensate to maintain  $V_{OUT}$  at the set level and noise on  $V_{OUT}$  will be reflected into the supply of each internal circuit. The reference voltage,  $V_{REF}$  is influenced by the  $C_{NOISE}$  pin. Noise into this pin will add to the reference voltage and be fed through the circuit. These factors will not cause a problem if some simple steps are taken. Figure 5 shows where these added ESR resistances are present in the typical LDO circuit.

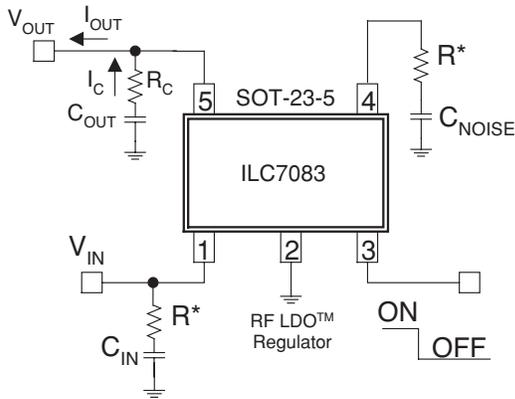


Figure 5. ESR Present in  $C_{OUT}$  and  $C_{NOISE}$

With this in mind low ESR components will offer better performance where the LDO may be subjected to large load transients current. ESR is less of a problem with  $C_{IN}$  as the voltage fluctuations at the input will be filtered by the LDO.

However, being aware of these current flows, there is also another potential source of induced voltage noise from the resistance inherent in the PCB trace. Figure 6 shows where the additive resistance of the PCB can manifest itself. Again

these resistances may be very small, but a summation of several currents can develop detectable voltage ripple and will be amplified by the LDO. Particularly the accumulation of current flows in the ground plane can develop significant voltages unless care is taken. With a degree of care, the ILC7083/ILC7084 will yield outstanding performance.

### Printed Circuit Board Layout Guidelines

As was mentioned in the previous section, to take full advantage of any high performance LDO regulator requires paying careful attention to grounding and printed circuit board (PCB) layout.

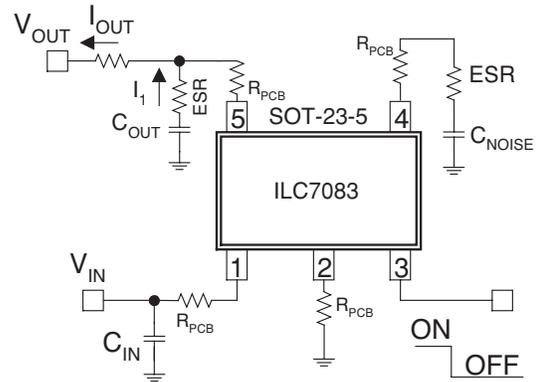


Figure 6. Inherent PCB resistance

Figure 7 shows the effects of poor grounding and PCB layout magnified by the ESR and PCB resistances and the accumulation of current flows.

Note that particularly during high output load current, the LDO regulator’s ground pin and the ground return for  $C_{OUT}$  and  $C_{NOISE}$  are not at the same potential as the system ground. This is due to high frequency impedance caused by PCB’s trace inductance and DC resistance. The current loop between  $C_{OUT}$ ,  $C_{NOISE}$  and the LDO regulator’s ground pin will degrade performance of the LDO.

Figure 8 shows an optimum schematic. In this schematic, high output surge current has little effect on the ground current and noise bypass current return of the LDO regulator. Note that the key difference here is that  $C_{OUT}$  and  $C_{NOISE}$  are directly connected to the LDO regulator’s ground pin. The LDO is then separately connected to the main ground plane and returned to a single point system ground.

The layout of the LDO and its external components are also based on some simple rules to minimize EMI and output voltage ripple.

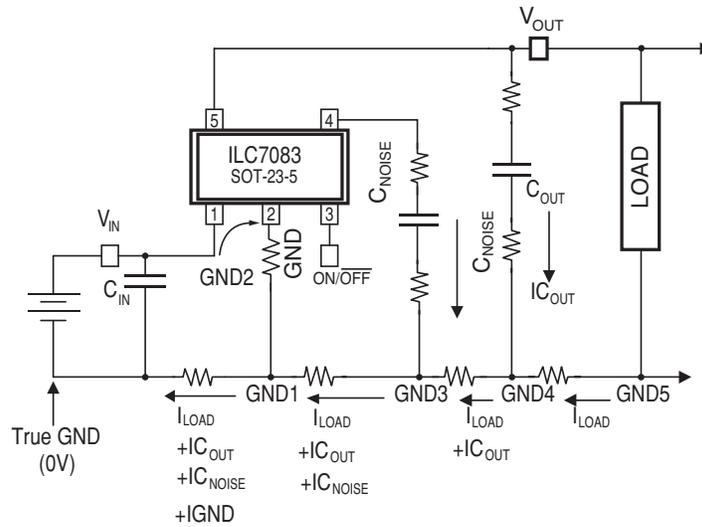


Figure 7. Effects of Poor Circuit Layout

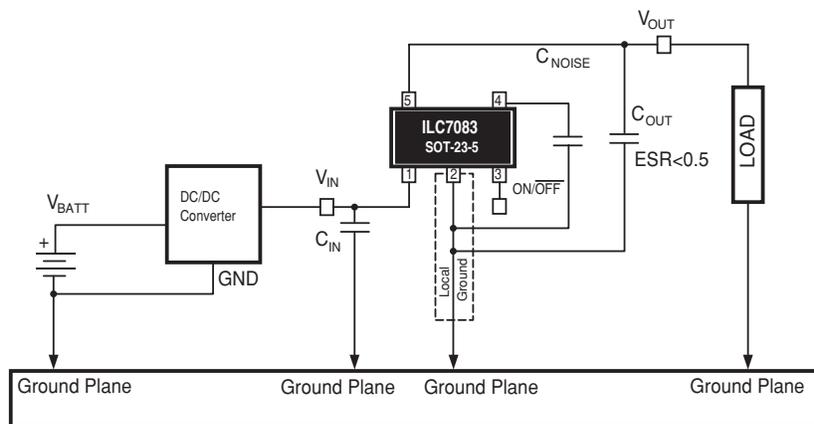


Figure 8. Recommended Application Circuit Schematic

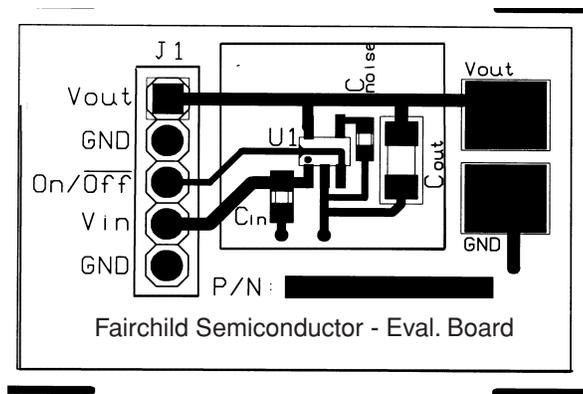


Figure 9. Recommended Application Circuit Layout (not drawn to scale)

Note, ground plane is bottom layer of PCB and connects to top layer ground connections through vias.

## Evaluation Board Parts List (for Figure 9)

Label	Part Number	Manufacturer	Description
U1	ILC7083AIM5-30	Fairchild Semiconductor	150mA RF LDO™ regulator
J1	69190-405	Berg	Connector, four position header
Cin	GRM40 Y5V 105Z16	muRata	Ceramic capacitor, 1μF, 16V, SMT (size 0805)
Cnoise	ECU-V1H103KBV	Panasonic	Ceramic capacitor, 0.01μF, 16V, SMT (size 0603)
Cout	GRM42-6X5R475K10	muRata	Ceramic capacitor, 4.7μF, 16V, SMT (size 1206)

### Grounding Recommendations

1. Connect  $C_{IN}$  between  $V_{IN}$  of the ILC7083 and the “GROUND PLANE”.
2. Keep the ground side of  $C_{OUT}$  and  $C_{NOISE}$  connected to the “LOCAL GROUND” and not directly to the “GROUND PLANE”.
3. On multilayer boards use component side copper for grounding around the ILC7083 and connect back to a “GROUND PLANE” using vias.
4. If using a DC-DC converter in your design, use a star grounding system with separate traces for the power ground and the control signals. The star should radiate from where the power supply enters the PCB.

### Layout Considerations

1. Place all RF LDO related components; ILC7083, input capacitor  $C_{IN}$ , noise bypass capacitor  $C_{NOISE}$  and output capacitor  $C_{OUT}$  as close together as possible.
2. Keep the output capacitor  $C_{OUT}$  as close to the ILC7083 as possible with very short traces to the  $V_{OUT}$  and GND pins.
3. The traces for the related components; ILC7083, input capacitor  $C_{IN}$ , noise bypass capacitor  $C_{NOISE}$  and output capacitor  $C_{OUT}$  can be run with minimum trace widths close to the LDO.
4. Maintain a separate “LOCAL GROUND” remote from the “GROUND PLANE” to ensure a quiet ground near the LDO.

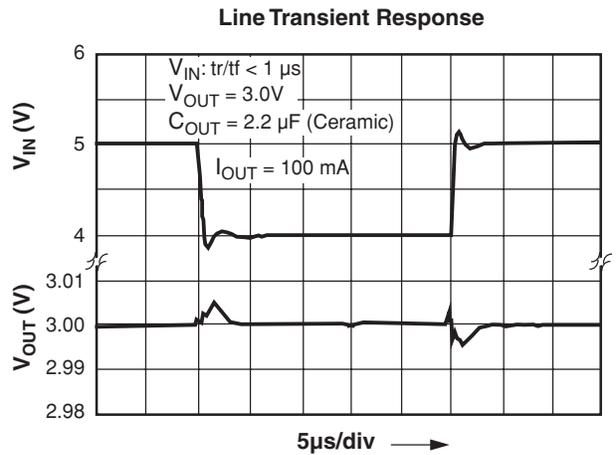
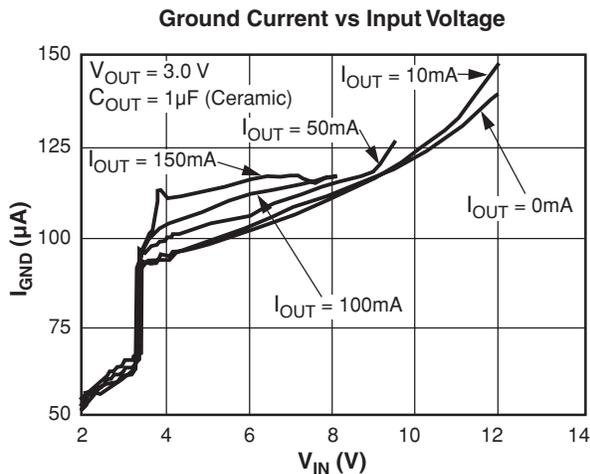
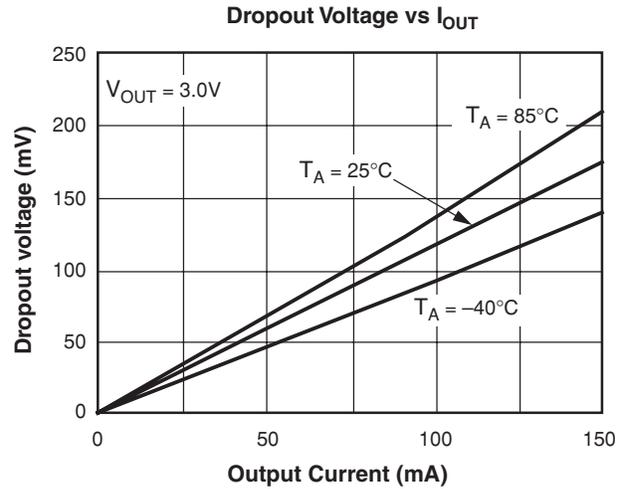
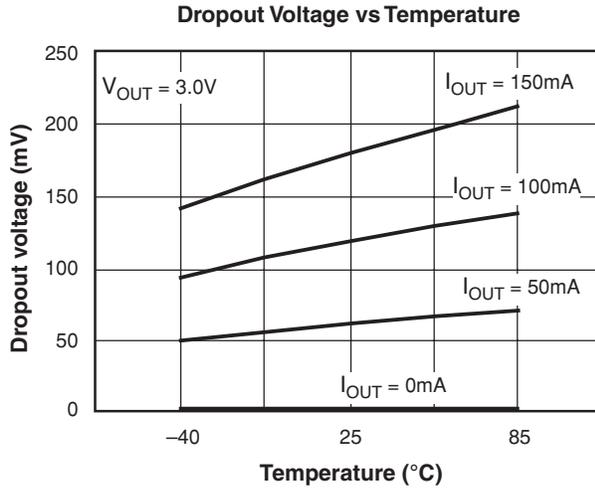
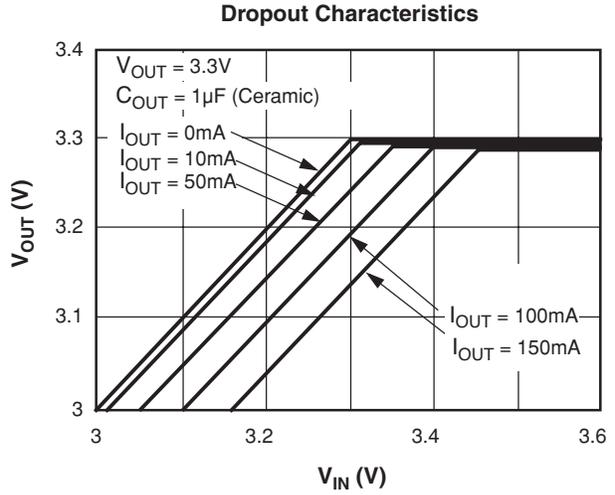
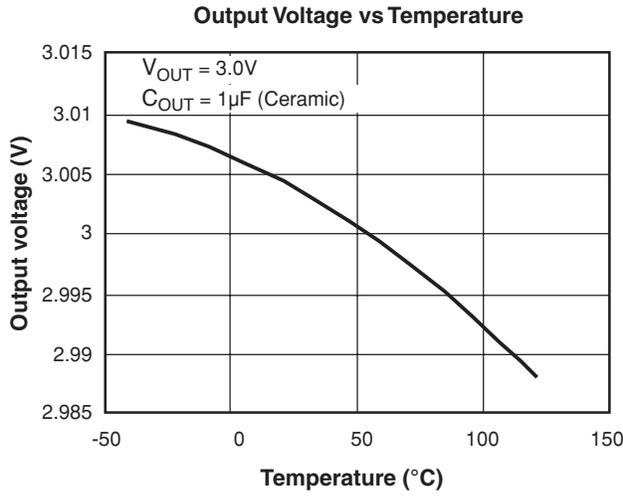
Figure 9 shows how this circuit can be translated into a PCB layout.

## Recommended Ceramic Output Capacitors

$C_{OUT}$	Capacitor Size	$I_{OUT}$	Dielectric	Part Number	Capacitor Vendor
1μF	0805	0 to 150mA	X5R	C2012X5R1A105KT	TDK
	0805		X7R	GRM40X7R105K010	muRata
	0805		X7R	LMK212BJ105KG	Taiyo-Yuden
	1206		X7R	GRM42-6X7R105K016	muRata
	1206		X7R	EMK316BJ105KL	Taiyo-Yuden
	1206		X5R	TMK316BJ105KL	Taiyo-Yuden
2.2μF	0805	0 to 150mA	X5R	GRM40X5R225K 6.3	muRata
	0805		X5R	C2012X5R0J225KT	TDK
	1206		X5R	EMK316BJ225ML	Taiyo-Yuden
4.7μF	1206	0 to 150mA	X5R	GRM42-6X5R475K010	muRata
	1206		X7R	LMK316BJ475ML	Taiyo-Yuden

## Typical Performance Characteristics ILC7083/ILC7084

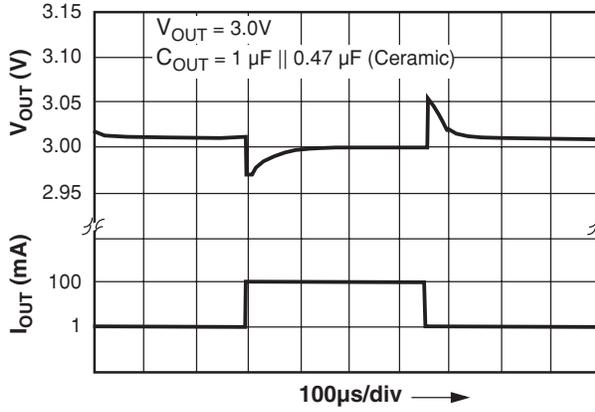
Unless otherwise specified:  $T_A = 25^\circ\text{C}$ ,  $V_{IN} = V_{OUT(NOM)} + 1\text{V}$ , ON/OFF pin tied to  $V_{IN}$



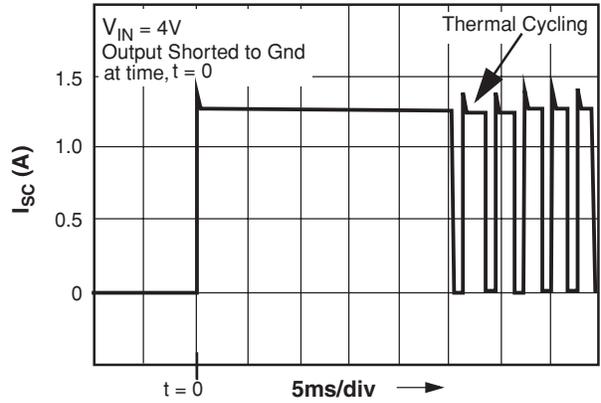
## Typical Performance Characteristics ILC7083/ILC7084

Unless otherwise specified:  $T_A = 25^\circ\text{C}$ ,  $V_{IN} = V_{OUT(NOM)} + 1\text{V}$ , ON/OFF pin tied to  $V_{IN}$

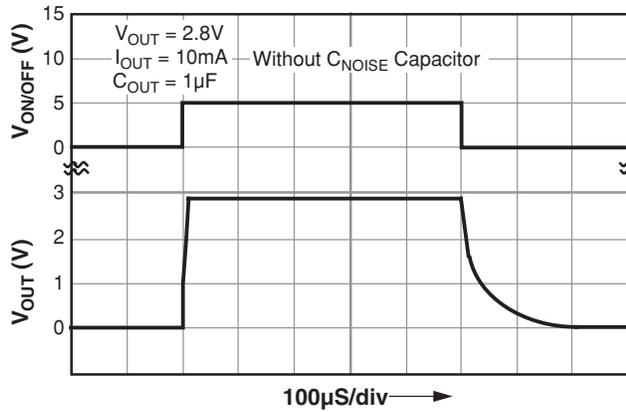
Load Transient Response



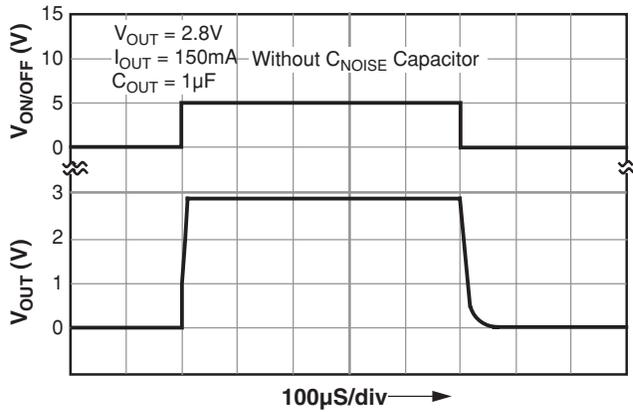
Short Circuit Current



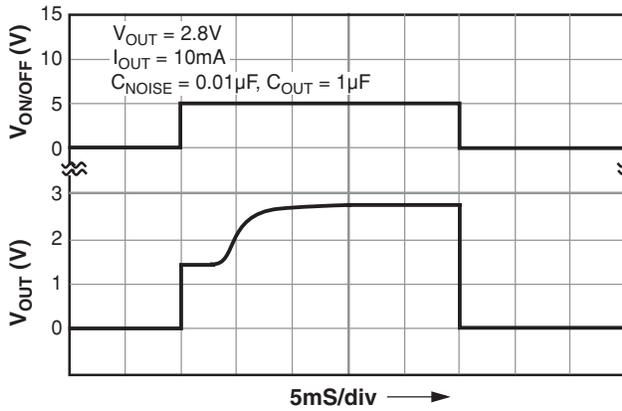
On/Off Transient Response



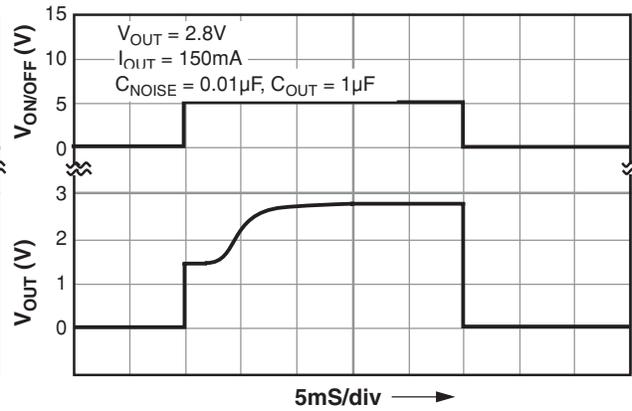
On/Off Transient Response



On/Off Transient Response

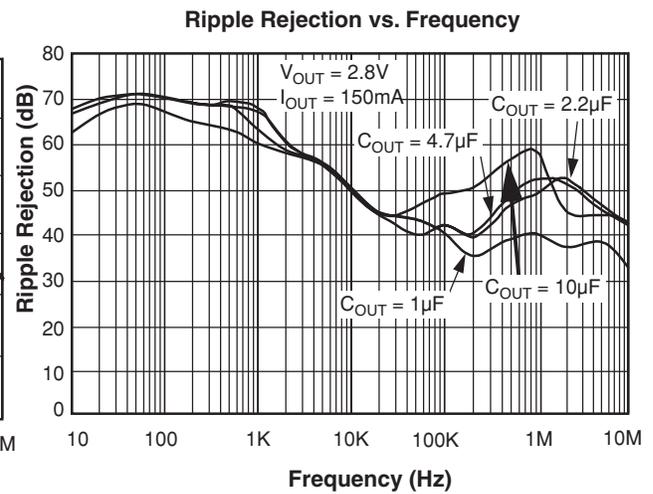
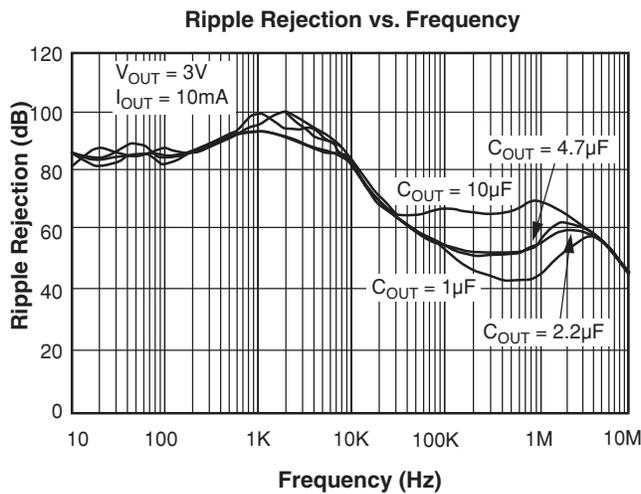
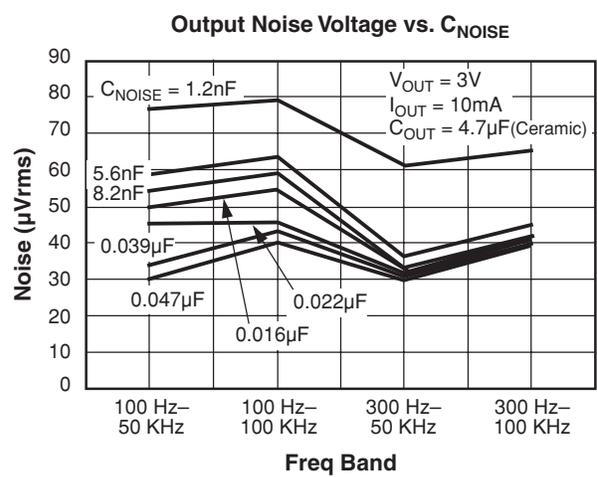
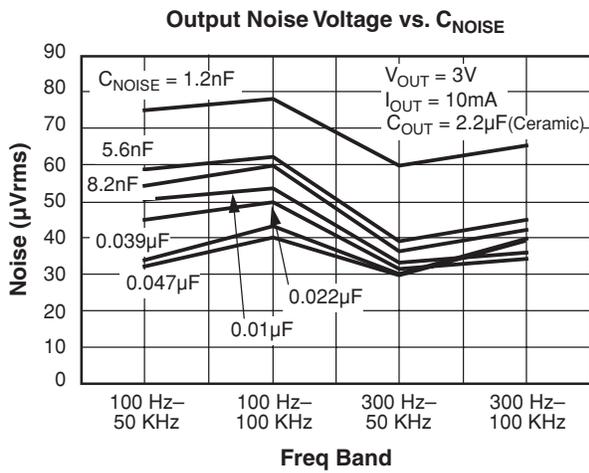
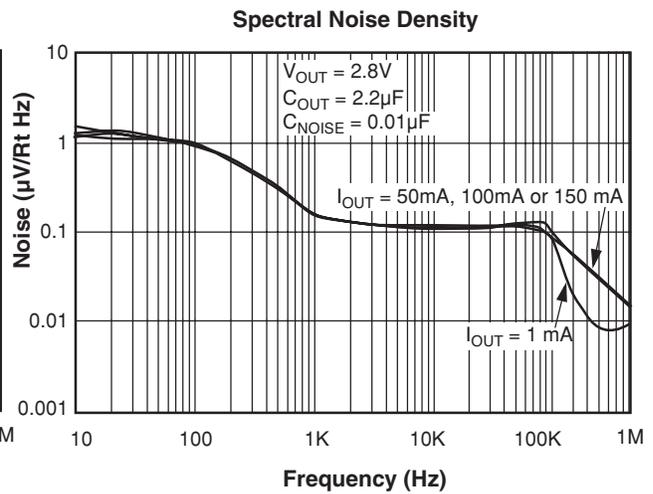
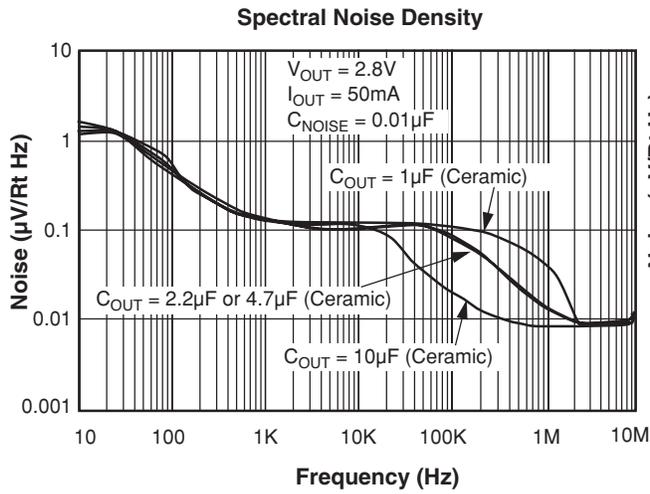


On/Off Transient Response



# Typical Performance Characteristics ILC7083/ILC7084

Unless otherwise specified:  $T_A = 25^\circ\text{C}$ ,  $V_{IN} = V_{OUT(NOM)} + 1\text{V}$ , ON/OFF pin tied to  $V_{IN}$



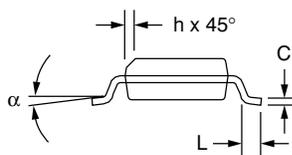
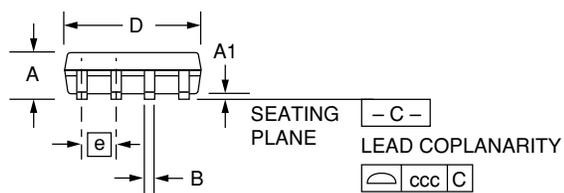
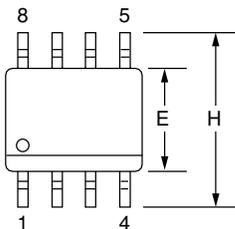
# Mechanical Dimensions

## 8-Lead Plastic Surface Mount (SOIC)

Symbol	Inches		Millimeters		Notes
	Min.	Max.	Min.	Max.	
A	.053	.069	1.35	1.75	
A1	.004	.010	0.10	0.25	
B	.013	.020	0.33	0.51	
C	.0075	.010	0.20	0.25	5
D	.189	.197	4.80	5.00	2
E	.150	.158	3.81	4.01	2
e	.050 BSC		1.27 BSC		
H	.228	.244	5.79	6.20	
h	.010	.020	0.25	0.50	
L	.016	.050	0.40	1.27	3
N	8		8		6
$\alpha$	0°	8°	0°	8°	
ccc	—	.004	—	0.10	

**Notes:**

1. Dimensioning and tolerancing per ANSI Y14.5M-1982.
2. "D" and "E" do not include mold flash. Mold flash or protrusions shall not exceed .010 inch (0.25mm).
3. "L" is the length of terminal for soldering to a substrate.
4. Terminal numbers are shown for reference only.
5. "C" dimension does not include solder finish thickness.
6. Symbol "N" is the maximum number of terminals.

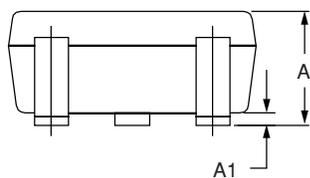
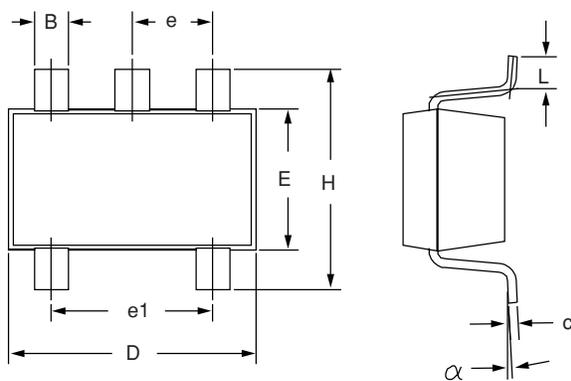


### 5-Lead Plastic Surface Mount (SOT-23-5)

Symbol	Inches		Millimeters		Notes
	Min.	Max.	Min.	Max.	
A	.035	.057	.90	1.45	
A1	.000	.006	.00	.15	
B	.008	.020	.20	.50	
c	.003	.010	.08	.25	
D	.106	.122	2.70	3.10	
E	.059	.071	1.50	1.80	
e	.037 BSC		.95 BSC		
e1	.075 BSC		1.90 BSC		
H	.087	.126	2.20	3.20	
L	.004	.024	.10	.60	
$\alpha$	0°	10°	0°	10°	

**Notes:**

1. Package outline exclusive of mold flash & metal burr.
2. Package outline exclusive of solder plating.
3. EIAJ Ref Number SC-74A.



## Ordering Information (T<sub>A</sub> = -40°C to +85°C)

Part Number	Output Voltage (V)	Package	Package Marking
ILC7083AIM525X	2.5	SOT-23-5	BM
ILC7083AIM527X	2.7	SOT-23-5	BN
ILC7083AIM5285X	2.85	SOT-23-5	BJ
ILC7083AIM530X	3.0	SOT-23-5	BB
ILC7083AIM533X	3.3	SOT-23-5	BC
ILC7083AIM536X	3.6	SOT-23-5	BD
ILC7083AIM550X	5.0	SOT-23-5	BE
ILC7083AIM5ADJX	Adjustable	SOT-23-5	BF
ILC7083AIK30X	3.0	SOIC-8	7083A1K3
ILC7083AIKADJ	Adjustable	SOIC-8	7083A1K
ILC7084AIM5ADJX	Adjustable	SOT-23-5	DF

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