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## Triple-Channel Video Reconstruction Filters with **Back-Porch Clamp to GND**

## **General Description**

The MAX7438/MAX7439 three-channel standard definition video reconstruction filters include a back-porch clamp that sets the output blanking level of the video signal to ground. Each channel of the MAX7438/ MAX7439 combines a lowpass filter with adjustable high-frequency boost levels and an output buffer capable of driving two standard 150 $\Omega$  video loads. The blanking level of the output video signal on each channel is clamped to ground, eliminating the need for large AC-coupling output capacitors. Direct input coupling circuitry eliminates the need for AC-coupling input capacitors. This DC-in/DC-out architecture results in extremely low line-time distortion. The MAX7438/ MAX7439 are ideal for antialiasing and DAC smoothing in digital video devices such as STBs, DVDs, PVRs, and hard disk recorders. The MAX7438/MAX7439 operate from ±5V dual supplies.

The three-channel MAX7438/MAX7439 are ideal for Y. Pb, Pr, and RGB component video signals, three composite video signals, and also Y/C plus CVBS video signals. Each filter channel achieves 60dB of attenuation at 27MHz and a maximally flat passband from DC to 5MHz.

The MAX7438 offers an internal gain of +2V/V, while the MAX7439 offers a gain of +3V/V.

#### **Applications**

Set-Top Boxes/HDR/DVD Game Consoles Camcorders

Composite, Component, S-Video Output for NTSC. PAL. SDTV

#### Features

- ♦ Back-Porch of Video Output Signal Clamped to Ground
- ♦ Eliminates Input/Output AC-Coupling Capacitors
- ♦ 0.1% Line-Time Distortion
- ♦ Stopband: 55dB at 27MHz
- ♦ Passband: ±0.8dB out to 5MHz
- ♦ Diff Gain = 0.05%, Diff Phase = 0.05 Degrees
- ♦ Output Clamped to Ground with Loss of Input
- ♦ Each Output Drives Two 150Ω Video Loads
- ♦ Up to 2dB of High-Frequency Boost Control
- ♦ Ideal for CVBS, Y/C (S-Video), and RGB (Y, Pb, Pr) Outputs for NTSC, PAL, and SDTV
- ♦ Filter Bypass Mode
- ♦ Small 20-Pin 5mm × 5mm Thin QFN Package

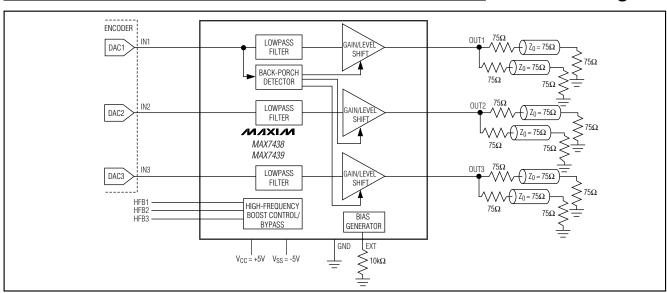
## **Ordering Information**

PART	TEMP RANGE	PIN-PACKAGE
MAX7438ETP	-40°C to +85°C	20 Thin QFN-EP*
MAX7439ETP	-40°C to +85°C	20 Thin QFN-EP*

<sup>\*</sup>EP = Exposed paddle

Pin Configuration appears at end of data sheet.

## Functional Diagram



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#### **ABSOLUTE MAXIMUM RATINGS**

$\begin{array}{cccccccccccccccccccccccccccccccccccc$	Operating Temperature Range40°C to +85°C Storage Temperature Range65°C to +150°C Junction Temperature+150°C Lead Temperature (soldering, 10s)+300°C
above +70 G)1000.7111W	

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

#### **ELECTRICAL CHARACTERISTICS**

 $(V_{CC} = +5V \pm 5\%, V_{SS} = -5V \pm 5\%, R_{EXT} = 10k\Omega \pm 1\%, R_{HFB\_-1} = 15k\Omega \pm 1\%, R_{HFB\_-2} = 1k\Omega \pm 1\%, no boost, R_{OUT} = 75\Omega, C_{OUT} = 0 to 20pF, T_A = T_{MIN} to T_{MAX}, unless otherwise noted. Typical values are at T_A = +25°C.)$ 

PARAMETER	SYMBOL	CONDIT	IONS	MIN	TYP	MAX	UNITS
Deschand Despense		f = 100kHz to 4.2MHz, r	no HF boost	-0.6		+0.6	dB
Passband Response		f = 100kHz to 5MHz, no	HF boost	-0.8		+0.8	иь
Stopband Attenuation	AsB	f = 27MHz, relative to 10	00kHz	40	60		dB
HF Boost Step Size		f = 4.2MHz			0.45		dB
		Bypass		0		50	
		No boost		280		360	
HED Voltage Bange	\/=n	Boost 1		670		850	mV
HFB_ Voltage Range	VHFB_	Boost 2		1360		1700	IIIV
		Boost 3		2250		2750	
		Boost 4		3500		Vcc	
Differential Cain	dG	5-step modulated	MAX7438, V <sub>IN</sub> = 1V <sub>P-P</sub>		0.05	0.5	0/
Differential Gain	dG	staircase	MAX7439, V <sub>IN</sub> = 670mV <sub>P-P</sub>		0.05	0.5	%
D''' I D		5-step modulated	MAX7438, V <sub>IN</sub> = 1V <sub>P-P</sub>		0.05	0.5	-
Differential Phase	dθ	staircase	MAX7439, V <sub>IN</sub> = 670mV <sub>P-P</sub>		0.05	0.5	Degrees
		f = 100kHz to 5MHz,	MAX7438, V <sub>IN</sub> = 700mV <sub>P-P</sub>		0.2	0.5	
Total Harmonic Distortion	THD	V <sub>IN</sub> = 0.7V <sub>P-P</sub>	MAX7439, V <sub>IN</sub> = 460mV <sub>P-P</sub>		0.2	0.5	%
Signal-to-Noise Ratio	SNR	Output signal (2V <sub>P-P</sub> ) to f = 100Hz to 5MHz	P-P noise,	68	75		dB
Group Delay Deviation	Δt <sub>G</sub>	Deviation from 100kHz t	to 3.58(4.43)MHz		12	30	ns
Group Delay Matching	tg(MATCH)	f = 100kHz			2		ns
Line-Time Distortion	HDIST	18µs, 100 IRE bar			0.1		%
Field-Time Distortion	V <sub>DIST</sub>	130 lines, 18µs, 100 IRE	bar		0.2		%

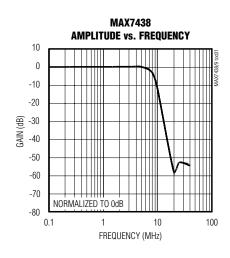
#### **ELECTRICAL CHARACTERISTICS (continued)**

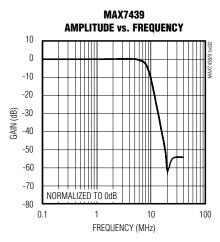
 $(V_{CC} = +5V \pm 5\%, V_{SS} = -5V \pm 5\%, R_{EXT} = 10k\Omega \pm 1\%, R_{HFB\_-1} = 15k\Omega \pm 1\%, R_{HFB\_-2} = 1k\Omega \pm 1\%, no boost, R_{OUT} = 75\Omega, C_{OUT} = 0 to 20pF, T_A = T_{MIN} to T_{MAX}, unless otherwise noted. Typical values are at T_A = +25°C.)$ 

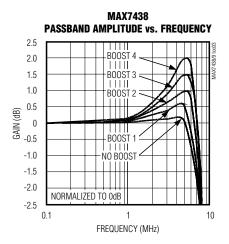
PARAMETER	SYMBOL	COND	DITIONS	MIN	TYP	MAX	UNITS
Clamp Accuracy		Output blanking level	relative to GND	-55		+55	mV
Clamp Settling Time		Back porch within the accuracy	e specified clamp			50	Lines
Law Farance Cain		0-1	MAX7438	5.5	6	6.5	-ID
Low-Frequency Gain		Gain at 100kHz	MAX7439	9	9.54	10	dB
Low-Frequency Gain Matching	tG	Channel-to-channel g	gain matching,			+5	%
Input Voltage Range		THD < 0.5%		-0.5		+1.8	V
Output Voltage Range	Vout	f = 5MHz		-0.8		+1.8	V
Channel-to-Channel Crosstalk	XTALK	Channel-to-channel c	rosstalk, f = 5MHz		-60		dB
Output Short-Circuit Current	I <sub>SC</sub>	OUT_ shorted to GNE	or V <sub>CC</sub>		50		mA
Input Leakage Current at IN_	I <sub>IN</sub> _				±1	±5	μΑ
Input Lookogo Current et UED	lues	Bypass mode, boost	1 to boost 3			0.2	
Input Leakage Current at HFB_	I <sub>HFB</sub> _	Boost 4				30	μΑ
Input Resistance				500			kΩ
Curantu Valtaria Dancia	Vcc			4.75		5.25	V
Supply Voltage Range	V <sub>SS</sub>			-4.75		-5.25	V
Cumply Current	Icc	$V_{CC} = +5.25V$ , no loa	d		110	160	m 1
Supply Current	I <sub>SS</sub>	$V_{SS} = -5.25V$ , no load	I		110	160	mA
Power-Supply Rejection Ratio	PSRR	$V_{IN} = 100 \text{mV}_{P-P}, f = 0$	) to 3.5MHz		30		dB

## Typical Operating Characteristics

 $(V_{CC} = +5V, V_{SS} = -5V, R_{OUT} = 150\Omega, R_{EXT} = 10k\Omega$ , no boost,  $T_A = +25^{\circ}C$ , unless otherwise noted.)

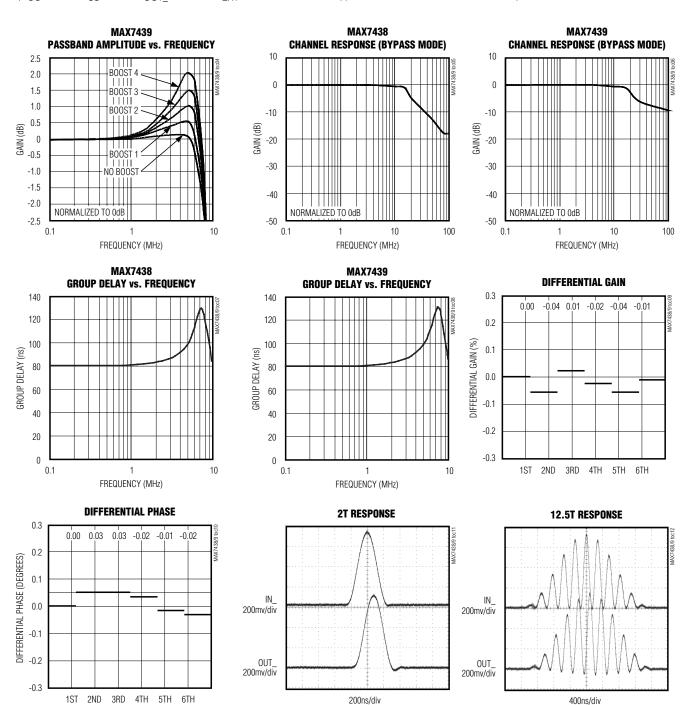






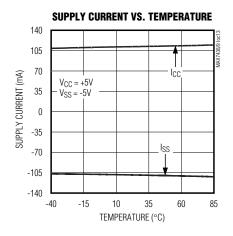
## Typical Operating Characteristics (continued)

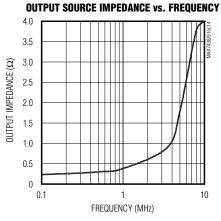
 $(V_{CC} = +5V, V_{SS} = -5V, R_{OUT} = 150\Omega, R_{EXT} = 10k\Omega, no boost, T_A = +25^{\circ}C, unless otherwise noted.)$ 

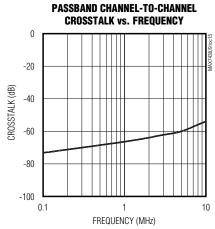


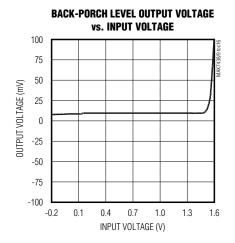
## **Typical Operating Characteristics (continued)**

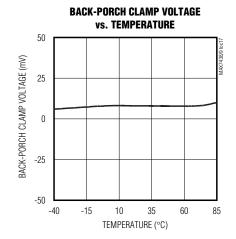
 $(V_{CC} = +5V, V_{SS} = -5V, R_{OUT} = 150\Omega, R_{EXT} = 10k\Omega, no boost, T_A = +25^{\circ}C, unless otherwise noted.)$ 











#### **Pin Description**

PIN	NAME	FUNCTION
1, 4, 7, 10, 11, 13, 15, 19	GND	Ground
2	IN1	Channel 1 Video Input
3	HFB1	Channel 1 High-Frequency Boost and Filter Bypass Control Input. Connect external resistors to HFB1 for high-frequency boost and filter bypass control. See Table 1.
5	OUT1	Channel 1 Video Output
6	Vcc	+5V Power Input
8	OUT2	Channel 2 Video Output
9	V <sub>SS</sub>	-5V Power Input. Connect the backside exposed pad to VSS.
12	OUT3	Channel 3 Video Output
14	HFB3	Channel 3 High-Frequency Boost and Filter Bypass Control Input. Connect external resistors to HFB3 for high-frequency boost and filter bypass control. See Table 1.
16	IN3	Channel 3 Video Input
17	HFB2	Channel 2 High-Frequency Boost and Filter Bypass Control Input. Connect external resistors to HFB2 for high-frequency boost and filter bypass control. See Table 1.
18	IN2	Channel 2 Video Input
20	EXT	External Bias Resistor. Connect a 10kΩ resistor from EXT to GND.
_	EP	Exposed Paddle. Connect to VSS. Do not connect to GND.

## Table 1. External Resistor Values for Bypass Mode and High-Frequency Boost Control

MODE	R <sub>HFB1</sub> (kΩ)	R <sub>HFB2</sub> (kΩ)	V <sub>HFB_</sub> (V) (V <sub>CC</sub> = 5V)	NOMINAL BOOST (dB)/ BYPASS (MAX7438)	NOMINAL BOOST (dB)/ BYPASS (MAX7439)
Bypass	Open	0	0	Bypass	Bypass
No boost	15	1	0.318	0	0
Boost 1	11.3	2	0.758	0.5	0.75
Boost 2	16.5	7.32	1.53	1.0	1.5
Boost 3	11.3	11.3	2.5	1.5	2.25
Boost 4	4.42	18.2	4.027	2.0	3.0

## **Detailed Description**

#### Filter

#### Filter Response

The MAX7438/MAX7439 reconstruction filters consist of three separate lowpass filters with Butterworth-type response. The filter features a maximally flat passband for NTSC and PAL bandwidths, while maintaining good group delay characteristics. The stopband offers excellent attenuation at frequencies of 27MHz and above (see the *Typical Operating Characteristics* section).

The autotrimming circuit digitally controls the corner frequency to maintain the frequency characteristics over process and temperature.

#### High-Frequency Boost

The high-frequency boost compensates for signal degradation and roll-off in the signal path prior to the MAX7438/MAX7439 to increase image sharpness. Program the level of high-frequency boost for each channel by selecting the corresponding external resistor values (RHFB\_-1 and RHFB\_-2, as shown in the *Typical Operating Circuit* section) given in Table 1. The

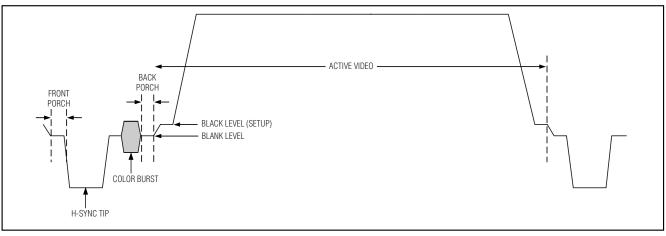


Figure 1. Standard Video Signal

external resistors form a voltage-divider between V<sub>CC</sub> and GND. The values in the fourth column (V<sub>HFB</sub> (V)) are calculated using the following equation:

$$V_{HFB} = V_{CC} \times R_{HFB}_{-2} / (R_{HFB}_{-1} + R_{HFB}_{-2})$$
  
where  $V_{CC} = 5V$ .

Apply an external voltage to HFB\_ in place of R<sub>HFB\_-1</sub> and R<sub>HFB\_-2</sub> as an alternative. See the *Electrical Characteristics* section. Filter boost modes can also be controlled from a microprocessor. See the *Applications Information* section.

#### Filter Bypass

The MAX7438/MAX7439 offer selectable filter bypassing that allows any of the video inputs to be filtered or unfiltered. Select the filter bypass mode for a given channel by setting the corresponding values for R<sub>HFB</sub> -1 and R<sub>HFB</sub> -2 according to Table 1.

#### **Output Buffer**

The output buffer is able to drive two standard  $150\Omega$  video loads with a 2V<sub>P-P</sub> signal. The MAX7438 output buffer has a preset gain of 2V/V, and the MAX7439 output buffer has a gain of 3V/V. The MAX7439 is ideal for a DAC output whose voltage range is between 0.67V and 1V. Set the DAC output to 0.67V to achieve a 2V/V signal on OUT\_.

#### **Back-Porch Clamp**

The MAX7438/MAX7439 feature a back-porch clamp to set the output blanking level. This clamp shifts the DC level of the video signal so that the back-porch level is close to ground (see Figure 1). The devices sense the voltage during back porch and feed back into a control

system that provides the appropriate DC-level shift in the filter channel to clamp the output to ground. The back-porch clamp to ground eliminates the need for large output-coupling capacitors that can introduce unwanted line-time distortion (tilt), cost, and board space. This feedback network and the on-chip capacitors introduce a finite settling time (50 lines max) after power-up or any dramatic shift in input voltage (see the *Electrical Characteristics* section).

Channel 1 requires a video signal with sync information (CVBS, Y, or G), since the other two channels are clamped from channel 1. In the absence of a sync on channel 1, the circuit forces all outputs actively and continuously to ground.

## Applications Information

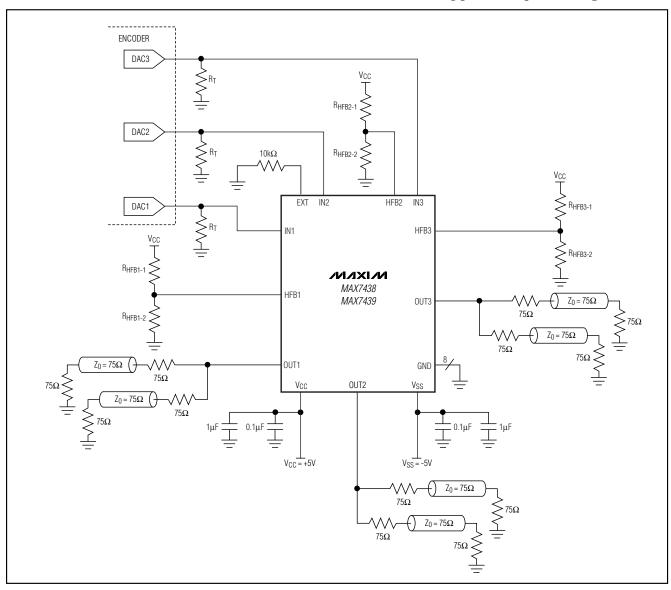
#### **Power-Supply Bypassing and Layout**

The MAX7438/MAX7439 operate from dual  $\pm 5V$  supplies. Bypass V<sub>CC</sub> and V<sub>SS</sub> to GND with  $0.1\mu F$  capacitors in parallel with  $1\mu F$  capacitors. Place the  $0.1\mu F$  capacitors as close to the power inputs as possible. Since EXT is a sensitive input, place R<sub>EXT</sub> close to the device to avoid signals coupling into EXT. Do not route any input, output, or dynamic signal near this pin and the accompanying trace.

**Note:** The exposed paddle is electrically connected to VSS.

Do not connect the exposed paddle to ground. Refer to the MAX7438 EV kit for layout examples, as well as a proven PC board layout example.

### **Typical Operating Circuit**



#### Microprocessor Control of High-Frequency Boost and Bypass

Use a DAC output to control the bypass and high-frequency boost levels on each channel (see Figure 2). Set the DAC output voltage to the corresponding bypass or boost levels desired (see Table 1).

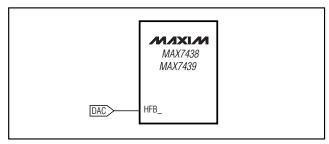


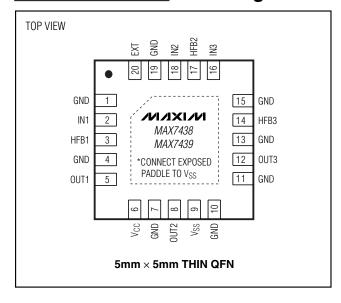
Figure 2. DAC Control of High-Frequency Boost and Bypass

## **Pin Configuration**

**Chip Information** 

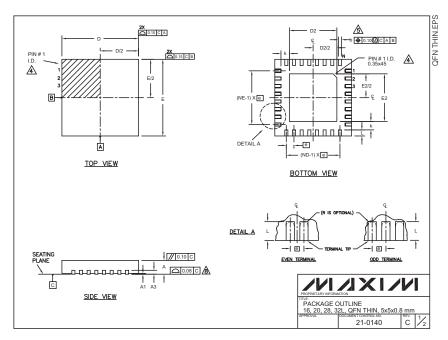
TRANSISTOR COUNT: 6418

PROCESS: BiCMOS



### Package Information

(The package drawing(s) in this data sheet may not reflect the most current specifications. For the latest package outline information, go to <a href="https://www.maxim-ic.com/packages">www.maxim-ic.com/packages</a>.)



				CC	OMMO	I DIME	NSIO	NS						EXPOS	ED P	AD V	AKIAI	IONS	•
PKG.		16L 5x5			20L 5x5		Т	28L 5x5	i		32L 5x5	$\overline{}$	PKG.	Т	D2			E2	
SYMBOL	MIN.	NOM.	MAX.	MIN.	NOM.	MAX.	MIN.	NOM.	MAX.	MIN.	NOM.	MAX.	CODES	MIN.	NOM.	MAX.	MIN.	NOM.	MAX.
Α	0.70	0.75	0.80	0.70	0.75	0.80	0.70	0.75	0.80	0.70	0.75	0.80	T1655-1	3.00	3.10	3.20	3.00	3.10	3.20
A1	0	0.02	0.05	0	0.02	0.05	0	0.02	0.05	0	0.02	0.05	T2055-2	3.00			3.00	3.10	
A3	- (	0.20 REF			).20 REF			0.20 RE	F.		0.20 REF	. 1	T2855-1	3.15		3.35		3.25	
b	0.25	0.30	0.35	0.25	0.30	0.35	0.20	0.25	0.30	0.20	0.25	0.30	T2855-2	2.60		2.80	2.60	2.70	2.80
D	4.90	5.00	5.10	4.90	5.00	5.10	4.90	5.00	5.10	4.90	5.00	5.10	T3255-2	3.00	3.10	3.20	3.00	3.10	3.20
Е	4.90	5.00	5.10	4.90	5.00	5.10	4.90	5.00	5.10	4.90	5.00	5.10							
0		0.80 BS	C.		0.65 BS	C.		0.50 BS	C.		0.50 BS	C.							
k	0.25	-	-	0.25	-	-	0.25	-	-	0.25	-	-							
L	0.45	0.55	0.65	0.45	0.55	0.65	0.45	0.55	0.65	0.30	0.40	0.50							
N		16			20			28			32								
ND		4			5			7			8								
NE		4			5			7			8								
NE		4 WHHB			5 WHHC			7 WHHD	-1		8 WHHD	-2							
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