# Linear Single Cell Li-Ion Battery Charger IC for Portable Applications

## **General Description**

The RT9536H is a fully integrated single cell Li-ion battery charger IC ideal for portable applications. The RT9536H optimizes the charging task by using a control algorithm including pre-charge mode, fast charge mode and constant voltage mode. The input voltage range of the VIN pin can be as high as 28V. When the input voltage exceeds the OVP threshold, it will turn off the charging MOSFET to avoid overheating of the chip.

In RT9536H, the maximum charging current can be programmed with an external resistor. For USB application, the user can set the current to 100mA/500mA through the EN/SET pin. For the factory mode, the RT9536H can allow 4.25V or 4.4V/2.3A power pass through to support system operation. It also provides a 50mA LDO to support the power of peripheral circuit. The internal thermal feedback circuit regulates the die temperature to optimize the charge rate for all ambient temperatures. The RT9536H provides protection functions such as under voltage protection, over voltage protection for VIN supply and thermal protection for battery temperature.

The RT9536H is available in a WDFN-10L 3x2 package to achieve optimized solution for PCB space and thermal considerations.

### Features

- 28V Maximum Rating for DC Adapter
- Internal Integrated Power MOSFETs
- Support 4.25V or 4.4V/2.3A Factory Mode
- 50mA Low Dropout Voltage Regulator
- Status Pin Indicator
- Programmed Charging Current
- Under Voltage Lockout
- Over Voltage Protection
- Thermal Feedback Optimized Charge Rate
- RoHS Compliant and Halogen Free

# Applications

- Cellular Phones
- Digital Cameras
- PDAs and Smart Phones
- Portable Instruments

## **Marking Information**

06=YM DNN 06= : Product Code YMDNN : Date Code

## **Simplified Application Circuit**





# **Ordering Information**

RT9536H口口

Package Type QW : WDFN-10L 3x2 (W-Type) Lead Plating System G : Green (Halogen Free and Pb Free)

Note :

Richtek products are :

- RoHS compliant and compatible with the current requirements of IPC/JEDEC J-STD-020.
- Suitable for use in SnPb or Pb-free soldering processes.

# **Functional Pin Description**

## **Pin Configurations**





WDFN-10L 3x2

Pin No.	Pin Name	Pin Function
1	VIN	Power Input.
2	ISET	Charging Current Setting.
3, 7, 11 (Exposed Pad)	GND	Ground. The exposed pad must be soldered to a large PCB and connected to GND for maximum power dissipation.
4	LDO	LDO Output (4.9V). This pin provides 50mA output current.
5	IEOC	End-of-Charge Current Setting. The $I_{\rm EOC}$ is from 5% to 50% Ichg-fast which is programmed by the ISET pin.
6	EN/SET	Enable and Operation Mode and VOUT Regulation Voltage Setting.
8	CHGSB	Indicator Output for Charging Status.
9	PGB	Indicator Output for Power Status.
10	BATT	Battery Charge Current Output.

# **Function Block Diagram**



## Operation

The RT9536H is designed for single cell Li-lon battery charger in portable applications.

#### **Base Circuit**

The Base circuit provides the internal power for VDD,  $V_{\text{REF}}$  reference voltage and bias current.

#### Power Switch with Switch Well Circuits

The main power switch between VIN and BATT is designed to control the charge current for battery. The switch well is designed to avoid the reverse current from battery to input power.

#### Sleep Mode

When the charger is only connected to battery with no input power, the charger will enter sleep mode and the leakage current from battery to the charger will be less than  $10\mu A$  for low power consumption

#### **Current Set Block**

The charge current is adjustable from the ISET pin with an external resistor between the ISET and GND. The Endof-Charge current is also adjustable by an external resistor connected from the IEOC pin to GND. If the charging current is less than EOC current, the CHGSB pin will be pulled high.

#### Protection

The protection circuits include OVP, UVLO and OTP. When the protection circuit is triggered, the main power switch will be turned of to protect the charging system.

#### Multi-Loop Controller

The Multi-Loop controller controls the operation during the charging process. The controller will make sure the battery is well charged in a suitable current, voltage, and die temperature.

#### **Status Indicator**

The CHGSB and PGB pin indicate the charger and power condition. During the charging process, the CHGSB pin is pulled low. When the charger is under charge done condition or abnormal condition, the CHGSB will be high impedance. The PGB pin indicates the input power status at VIN pin. When the input power is normal, the PGB pin is pulled low.

#### **Operation Mode**

The RT9536H provides programmable output current mode setting including USB100, USB500 and Factory mode. The operation mode is programmable through the input pulse number at EN/SET pin. Under the factory mode, besides the EN/SET is also used to set regulation to be 4.25V or 4.4V. Output current is up to 2.3A.

#### LDO

The RT9536H provides a LDO regulator to support the peripheral circuits. The output voltage is regulated to 4.9V and the maximum output current is 50mA.



# Absolute Maximum Ratings (Note 1)

Supply Input Voltage, VIN	0.3V to 28V
Other Pins	0.3V to 6V
• Power Dissipation, $P_D @ T_A = 25^{\circ}C$	
WDFN-10L 3x2	- 2.27W
Package Thermal Resistance (Note 2)	
WDFN-10L 3x2, $\theta_{JA}$	- 44°C/W
WDFN-10L 3x2, $\theta_{JC}$	- 11°C/W
Lead Temperature (Soldering, 10 sec.)	- 260°C
Junction Temperature	- 150°C
Storage Temperature Range	65°C to 150°C

## Recommended Operating Conditions (Note 3)

•	Supply Input Voltage, VIN	4.3V to 5.5V
•	Junction Temperature Range	$-40^{\circ}C$ to $125^{\circ}C$
•	Ambient Temperature Range	–20°C to 85°C

## **Electrical Characteristics**

(V<sub>IN</sub> = 5V, V<sub>BATT</sub> = 4V,  $T_A$  = 25°C, unless otherwise specified)

Parameter	Symbol	Test Conditions	Min	Тур	Max	Unit
VIN POR Rising Threshold Voltage	VPOR		3.15	3.3	3.45	V
VIN POR Threshold Voltage Hysteresis				200	300	mV
VIN OVP Threshold Voltage	Vovp		6.7	6.9	7.1	V
VIN OVP Threshold Voltage Hysteresis				200	300	mV
VIN – VOUT VOS Rising				75	150	mV
VIN – VOUT VOS Falling			18	32		mV
VIN Standby Current		V <sub>BATT</sub> = 4.5V, EN/SET = High		250	300	μΑ
VIN Supply Current		$V_{BATT} = 4.5V, EN/SET = Low$		1	2	mA
VOUT Sleep Leakage Current				1	10	μA
	Maria		4.186	4.25	4.313	V
VOUT Regulation	Vcv	$0^{\circ}$ C to $85^{\circ}$ C, $I_{LOAD} = 0$ mA	4.356	4.4	4.444	V
Thermal Regulation				125		°C
OTP				155		°C
OTP Hysteresis				20		°C
PGB/CHGSB Sink Current			20			mA
Dro Chorgo Throphold		$V_{OUT}$ Rising; $V_{CV}$ = 4.25V	2.4	2.5	2.6	v
Pre-Charge Threshold		$V_{OUT}$ Rising; $V_{CV} = 4.4V$	2.5	2.6	2.7	V
Fast-Charge to Pre-Charge Deglitch Time				25		ms

Paramet	ter	Symbol	Test Conditions	Min	Тур	Max	Unit	
Pre-Charge Current		IPRECHG	USB100 Mode	90	95	100	mA	
			USB500 Mode or ISET Mode, ratio of fast-charge current	15	20	25	%	
End of Charge Curr	rent (EOC)				REOC / KEOC		%	
IEOC Setting Curre	ent	IEOC		70	75	80	μA	
IEOC Setting KEOC				180	200	220	Ω/%	
VIN Power FET RD	S(ON)		Iout = 1A		280	512	mV	
ISET Set Voltage		VISET			1.5		V	
ISET Short Protect	Threshold			320		460	Ω	
ISET Short Protect Time	Deglitch				1.5		ms	
ISET Short Protect Current	Maximum				2		А	
			As ISET Mode, R <sub>ISET</sub> = 530	0.9	1	1.1	А	
VIN Charge Curren	t	ICHRG	As USB100 Mode	90	95	100	mA	
			As USB500 Mode	380	395	415	mA	
EN/SET Pull Low R	Resistor				200		kΩ	
EN/SET Voltage	Logic-High	VIH		1.4			v	
EN/SET VOILage	Logic-Low	VIL				0.4		
LDO On-Resistance	e	R <sub>DS(ON)</sub>			3	6	Ω	
LDO Output Voltag	е	Vldo		4.75	4.9	5.05	V	
LDO Maximum Out	tput Current			60	120	180	mA	
Factory Mode Vou	F		Vcv = 4.25V	4.165	4.25	4.335	V	
Factory Mode Vout			$V_{CV} = 4.4V$	4.312	4.4	4.488		
Factory Mode Maximum Output Current				2.3			А	
EN/SET Off Time			Timer to disable chip	2			ms	
EN/SET Lock Time			Timer to lock pulse count	1.5			ms	
EN/SET Duration	Logic-High			100		700		
	Logic-Low			100		700	μS	
EN/SET Set Time			Timer to set $V_{CV} = 4.4V$	750		1000	μS	

- Note 1. Stresses beyond those listed "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions may affect device reliability.
- Note 2.  $\theta_{JA}$  is measured at  $T_A = 25^{\circ}C$  on a high effective thermal conductivity four-layer test board per JEDEC 51-7.  $\theta_{JC}$  is measured at the exposed pad of the package.
- Note 3. The device is not guaranteed to function outside its operating conditions.



# **Typical Application Circuit**



# **Typical Operating Characteristics**







VOUT Sleep Leakage Current vs. Battery Voltage



LDO Voltage vs. Temperature















# **Application Information**

#### Description

The RT9536H is a fully integrated low cost single-cell Lilon battery charger IC with a Constant Current mode (CC mode) or a Constant Voltage mode (CV mode). The charge current is programmable to USB100, USB500 or ISET mode and the CV mode voltage is fixed at 4.25V/4.4V. The pre-charge threshold is fixed at 2.5V/2.64V. If the battery voltage is below the pre-charge threshold, the RT9536H charges the battery with a trickle current until the battery voltage rises above the pre-charge threshold. The RT9536H is capable of being powered up from AC adapter and USB (Universal Serial Bus) port inputs. Moreover, the RT9536H include a linear regulator (LDO 4.9V, 50mA) for supplying low power external circuitry.

#### **ACIN Over Voltage Protection**

The input voltage is monitored by the internal comparator and the input over voltage protection threshold is set to 6.9V. However, input voltage over 28V will still cause damage to the RT9536H. When the input voltage exceeds the threshold, the comparator outputs a logic signal to turn off the power P-MOSFET to prevent the high input voltage from damaging the electronics in the handheld system. When the input over voltage condition is removed, the comparator re-enables the output by running through the soft-start.

#### Charger Enable and mode Setting

EN/SET is used to enable or disable the charger as well as to select the charge current limit. Drive the EN pin to low or leave it floating to enable the charger. The EN/SET pin has a 200k $\Omega$  internal pull down resistor. So, when left floating, the input is equivalent to logic low. Drive this pin to high to disable the charger. After the EN/SET pin pulls low for 50µs, the RT9536H enters the USB500 mode and wait for the setting current signal. EN/SET can be used to program the charge current during this cycle. The RT9536H will change its charge current by sending different pulse to EN/SET pin. If no signal is sent to EN/SET, the RT9536H will remain in USB500 mode. A correct period of time for high pulse is between 100µs and 700µs and



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A. USB500 Mode with CV = 4.25V



B. ISET Mode with CV = 4.25V



C. USB100 Mode with CV = 4.25V



D. PTM Mode with CV = 4.25V



 $100\mu s < t_{High} < 700\mu s$ 

 $100\mu s < t_{Low} < 700\mu s$ 

E. USB500 Mode with CV = 4.4V



Four pulses to set USB500 mode Besides, wait the time,  $t_{Wait}$  and send a pulse of  $t_{CV\_SET}$  width for CV=4.4V

F. ISET Mode with CV = 4.4V



One pulse to set ISET mode Besides, wait the time,  $t_{Wait}$  and send a pulse of  $t_{CV\_SET}$  width for CV = 4.4V

G. USB100 Mode with CV = 4.4V



Two pulses to set USB100 mode Besides, wait the time,  $t_{Wait}$  and send a pulse of  $t_{CV\_SET}$  width for CV = 4.4V

H. PTM Mode with CV = 4.4V



Three pulses to set PTM mode Besides, wait the time,  $t_{Wait}\,and\,send\,a$  pulse of  $t_{CV\_SET}$  width for CV=4.4V

 $100\mu s < t_{High} < 700\mu s, 100\mu s < t_{Low} < 700\mu s, 1.5ms < t_{Wait} 750\mu s < t_{CV_Set} < 1ms$ 

Figure 1 (b). CV is 4.4V

Pulses	Charge Condition	MODE Control	
0	USB500 Mode	Charge Current Limit	
1	ISET Mode	Charge Current Limit	
2	USB100 Mode	Charge Current Limit	
3	Factory Mode	Enabled	
≥4	USB500 Mode	Charge Current Limit	

#### Table 1. Pulse Counting Map for EN/SET Interface

# RT9536H



#### **Battery Charge Profile**

The RT9536H charges a Li-Ion battery with a Constant Current (CC) or a Constant Voltage (CV).

The constant current is decided by the operation mode of USB100, USB500 or ISET mode. The constant current is set with the external resistor  $R_{\text{ISET}}$  and the constant voltage is fixed at 4.25V/4.4V. If the battery voltage is below the Pre-Charge Threshold, the RT9536H charges the battery with a trickle current until the battery voltage rises above the trickle charge threshold. When the battery voltage reaches 4.25V/4.4V, the charger enters CV mode and regulates the battery voltage at 4.25V/4.4V to fully charge the battery without the risk of over charging





#### **Battery Pre-Charge Current**

During a charge cycle, if the battery voltage is below the pre-charge threshold, the RT9536H enters the pre-charge mode. This feature revives deeply discharged cells and protects battery. Under USB100 Mode, the pre-charge current is internally set to 95mA. When the RT9536H is under USB500 and ISET Mode, the pre-charge current is 20% of fast-charge current set by external resistor R<sub>ISET</sub>.

#### **Battery Fast-Charge Current**

#### **ISET Mode**

The RT9536H offers ISET pin to program the charge current. The resistor  $R_{ISET}$  is connected to ISET and GND. The parameter  $K_{ISET}$  is specified in the specification Table.



#### USB500 and USB100 Mode

The fast-charge current is 95mA in USB100 mode and 395mA in USB500 mode. Note that if the fast-charge current set by external resistor is smaller than that in USB500 mode (395mA), the RT9536H charges the battery in ISET mode.

#### Battery Voltage Regulation (CV Mode)

The battery voltage regulation feedback is through the BATT pin. The RT9536H monitors the battery voltage between BATT and GND pins. When the battery voltage closes in on the battery regulation voltage threshold, the voltage regulation phase begins and the charging current begins to taper down. When the charging current falls below the programmed end-of-charge current threshold, the CHGSB pin goes high to indicate the termination of charge cycle.

The end-of-charge current threshold is set by the IEOC pin. The resistor  $R_{EOC}$  is connected to IEOC and GND. The parameters  $K_{EOC}$  and IEOC are specified in the specification Table.

 $I_{\text{EOC}}(\%) = \frac{R_{\text{EOC}}}{K_{\text{EOC}}} \ ; \ K_{\text{EOC}} = 200$ 

The current threshold of IEOC (%) is defined as the percentage of fast-charge current set by  $R_{ISET}$ . After the CHGSB pin is pulled high, the RT9536H still monitors the battery voltage. Charge current is resumed when the battery voltage goes to lower than the battery regulation voltage threshold.

#### **Factory Mode**

The RT9536H provides factory mode for supplies up to 2.3A for powering external loads with no battery installed and BATT is regulated to 4.25V. The factory mode allows the user to supply system power with no battery connected. In factory mode, thermal regulation is disabled but thermal protection (155°C) is still active. When using currents greater than 1.5A in factory mode, the user must limit the duty cycle at the maximum current to 20% with a maximum period of 10ms.

#### LDO

The RT9536H integrates one low dropout linear regulator (LDO) that supplies up to 50mA. The LDO is active whenever the input voltage is between POR threshold and OVP threshold. It is not affected by the EN/SET input. Note that the LDO current is independence and not monitored by the charge current limit.

#### Charge Status Outputs (CHGSB and PGB)

The open-drain CHGSB and PGB outputs indicate various charger operations as shown in the following Table. These status pins can be used to drive LEDs or communicate to the host processor. Note that ON indicates the open-drain transistor is turned on and LED is bright.

Condition	CHGSB	PGB
Input OVP	OFF	OFF
Input UVLO	OFF	OFF
Charge (CC Mode and CV Mode)	ON	ON
Charge Done (IFULL)	OFF	ON

	PGB Deglitches Time			
Condition	EN/SET is High	EN/SET is Low		
Entering OVP (V <sub>IN</sub> = 5.5V→10V)	0	100µs		
Leaving OVP (V <sub>IN</sub> = 10V→5.5V)	500µs	450µs		
Entering SLEEP $(V_{IN} = 5.5V \rightarrow 3.6V)$	0	32ms		
Leaving SLEEP (V <sub>IN</sub> = 3.6V→5.5V)	500µs	500µs		
Entering UVLO (V <sub>IN</sub> = 5.5V→2.5V)	0	0		
Leaving UVLO (V <sub>IN</sub> = 2.5V→5.5V)	230µs	230µs		

#### Sleep Mode

The RT9536H enters sleep mode if the power is removed from the input. This feature prevents draining the battery during the absence of input supply.

#### **Temperature Regulation and Thermal Protection**

In order to maximize charge rate, the RT9536H features a junction temperature regulation loop. If the power dissipation of the IC results in a junction temperature greater than the thermal regulation threshold ( $125^{\circ}C$ ), the RT9536H limits the charge current in order to maintain a junction temperature around the thermal regulation threshold ( $125^{\circ}C$ ). The RT9536H monitors the junction temperature, T<sub>J</sub>, of the die and disconnects the battery from the input if T<sub>J</sub> exceeds  $125^{\circ}C$ . This operation continues until junction temperature falls below thermal regulation threshold ( $125^{\circ}C$ ) by the hysteresis level. This feature prevents maximum power dissipation from exceeding typical design conditions.

#### Selecting the Input and Output Capacitors

In most applications, all that is needed is a high-frequency decoupling capacitor on the input. A 1 $\mu$ F ceramic capacitor, placed in close proximity to input to GND, works well. In some applications depending on the power supply characteristics and cable length, it may be necessary to add an additional 10 $\mu$ F ceramic capacitor to the input. The RT9536H requires a small output capacitor for loop stability. A typical 1 $\mu$ F ceramic capacitor placed between the BATT pin and GND is sufficient.

#### **Thermal Considerations**

For continuous operation, do not exceed absolute maximum operation junction temperature. The maximum power dissipation depends on thermal resistance of the IC package, PCB layout, rate of surrounding airflow, and difference between junction and ambient temperature. The maximum power dissipation can be calculated by the following formula :

 $\mathsf{P}_{\mathsf{D}(\mathsf{MAX})} = \left(\mathsf{T}_{\mathsf{J}(\mathsf{MAX})} - \mathsf{T}_{\mathsf{A}}\right) / \, \theta_{\mathsf{JA}}$ 

where  $T_{J(MAX)}$  is the maximum operation junction temperature,  $T_A$  is the ambient temperature, and  $\theta_{JA}$  is the junction to ambient thermal resistance.

For recommended operating condition specifications, the maximum junction temperature is 125°C. The junction to ambient thermal resistance,  $\theta_{JA}$ , is layout dependent. For WDFN-10L 3x2 package, the thermal resistance,  $\theta_{JA}$ , is 44°C/W on a standard JEDEC 51-7 four-layer thermal test board. The maximum power dissipation at  $T_A = 25$ °C can be calculated by the following formula :

 $P_{D(MAX)}$  = (125°C - 25°C) / (44°C/W) = 2.27W for WDFN-10L 3x2 package

The maximum power dissipation depends on the operating ambient temperature for fixed  $T_{J(MAX)}$  and thermal resistance,  $\theta_{JA}$ . The derating curve in Figure 4 allows the designer to see the effect of rising ambient temperature on the maximum power dissipation.



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#### **Layout Consideration**

The RT9536H is a fully integrated low cost single-cell Lilon battery charger IC ideal for portable applications. Careful PCB layout is necessary. For best performance, place all peripheral components as close to the IC as possible. A short connection is highly recommended. The following guidelines should be strictly followed when designing a PCB layout for the RT9536H.

- Input capacitor should be placed close to the IC and connected to ground plane. The trace of input in the PCB should be placed far away from the sensitive devices or shielded by the ground.
- The GND should be connected to a strong ground plane for heat sinking and noise protection.
- The connection of R<sub>ISET</sub> and R<sub>IEOC</sub> should be isolated from other noisy traces. The short wire is recommended to prevent EMI and noise coupling.
- Output capacitor should be placed close to the IC and connected to ground plane to reduce noise coupling.



Figure 5. Layout Guide

# **Outline Dimension**





DETAIL A Pin #1 ID and Tie Bar Mark Options

Note : The configuration of the Pin #1 identifier is optional, but must be located within the zone indicated.

Symbol	Dimensions	n Millimeters	<b>Dimensions In Inches</b>		
Symbol	Min	Max	Min	Max	
А	0.700	0.800	0.028	0.031	
A1	0.000	0.050	0.000	0.002	
A3	0.175	0.250	0.007	0.010	
b	0.180	0.300	0.007	0.012	
D	2.900	3.100	0.114	0.122	
D2	2.450	2.550	0.096	0.100	
E	1.900	2.100	0.075	0.083	
E2	0.750	0.850	0.030	0.033	
е	0.500		0.0	)20	
L	0.250	0.350	0.010	0.014	

W-Type 10L DFN 3x2 Package

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