# **General Description**

The MAX1749 allows for a constant vibration force while operating from a +2.5V to +6.5V input range and delivering up to 120mA. A PMOS pass transistor allows the 80µA supply current to remain independent of the load.

The output voltage can be adjusted from +1.25V to VIN with an external resistor-divider. When turned off (ON = low), the MAX1749 supply current drops to  $1\mu$ A (max) to minimize battery drain. Other features include short-circuit protection, thermal shutdown protection, and reverse battery protection. The MAX1749 is available in a 5-pin SOT23 package.

### Features

- Fixed 1.25V or Adjustable (1.25V to 6.5V) Output
- Low Cost
- Thermal Overload Protection
- Output Current Limit
- Reverse Battery Protection
- Low 0.1nA Off Supply Current
- Low 80µA Full-Load Supply Current

### Applications

Wireless Handset Vibrator Motor Drivers Pager Vibrator Motor Drivers

PART	TEMP. RANGE	PIN- PACKAGE	TOP MARK
/IAX1749EUK-T	-40°C to +85°C	5 SOT23	ADIX

#### 1 25V UP TO 120mA 2.5V TO 6.5V OUT IN COUT $C_{\text{IN}}$ ΜΛΧΙΜ 1μF 1μF ON MAX1749 ON SET BATTERY OFF GND

# **Typical Operating Circuit**



### M/X/M

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# **Ordering Information**

### **ABSOLUTE MAXIMUM RATINGS**

IN to GND	7V to +7V
Output Short-Circuit Duration	Infinite
SET to GND	0.3V to +7V
ON to GND	7V to +7V
ON to IN	7V to +0.3V
OUT to GND	0.3V to (V <sub>IN</sub> + 0.3V)
Continuous Power Dissipation ( $T_A = +7$	70°C)
SOT23-5 (derate 7.1mW/°C above -	

Operating Temperature Range	40°C to +85°C
Junction Temperature	+150°C
θJA	+140°C/W
Storage Temperature Range	
Lead Temperature (soldering, 10sec)	+300°C

Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.

### **ELECTRICAL CHARACTERISTICS**

( $V_{IN}$  = +3.6V,  $T_A$  = -40°C to +85°C, unless otherwise noted. Typical values are at  $T_A$  = +25°C.) (Note 1)

PARAMETER	SYMBOL	CONE	DITIONS	MIN	TYP	MAX	UNITS
Input Voltage	VIN			2.5		6.5	V
Output Voltage	Vout	$0.1 \text{mA} \le I_{OUT} \le 50 \text{mA}$	, SET = OUT	1.13	1.25	1.37	V
Maximum Output Current				120			mA
Current Limit (Note 2)	ILIM				280		mA
Ground Pin Current	lq				80	200	μA
Off Supply Current	IOFF	V <sub>OUT</sub> = V <sub>ON</sub> = GND	$T_A = +25^{\circ}C$		0.0001	1	- μΑ
			$T_A = +85^{\circ}C$		0.02		
ON INPUT			-				
ON Input High Threshold	VIH			2.0			V
ON Input Low Threshold	VIL					0.4	V
ON Input Bias Current	ION	ON = IN	$T_A = +25^{\circ}C$		0	100	nA
			$T_A = +85^{\circ}C$		0.05		
SET INPUT				I			
SET Input Leakage Current	I <sub>SET</sub>	$V_{SET} = 1.4V$	$T_{A} = +25^{\circ}C$		0.03	10	nA
			$T_A = +85^{\circ}C$		0.5		
THERMAL PROTECTION	•						
Thermal Shutdown Temperature	TSHDN				170		°C
Thermal Shutdown Hysteresis	ΔTSHDN				20		°C

Note 1: Limits are 100% production tested at T<sub>A</sub> = +25°C. Limits over the operating temperature range are guaranteed through correlation using Statistical Quality Control (SQC) methods.

Note 2: Not tested. For design purposes, the current limit should be considered 120mA minimum to 420mA maximum.



#### **Typical Operating Characteristics** (VIN = +3.6V, SET = OUT, CIN = 1µF, COUT = 1µF, TA = +25°C, unless otherwise noted.) **OUTPUT VOLTAGE OUTPUT VOLTAGE** vs. LOAD CURRENT vs. INPUT VOLTAGE **SUPPLY CURRENT vs. LOAD CURRENT** 1.4 1.27 100 90 1.2 1.26 80 SUPPLY CURRENT (µA) OUTPUT VOLTAGE (V) 70 1.25 60 1.24 50 40 1.23 30 20 1.22 0.2 10 0 1.21 0 2 5 6 0 1 3 4 0 20 40 100 120 60 80 0 20 40 60 80 100 120 INPUT VOLTAGE (V) LOAD CURRENT (mA) LOAD CURRENT (mA) SUPPLY CURRENT **OUTPUT VOLTAGE** vs. INPUT VOLTAGE vs. TEMPERATURE 120 1.28 $I_{LOAD} = 50 \text{mA}$ 1.27 100 SUPPLY CURRENT (µA) LOAD = 50 mAOUTPUT VOLTAGE (V) 1.26 80 1.25 60 $I_{LOAD} = 0$ 1.24 40 1.23 20 1.22 0 1.21 0 1 2 3 4 5 6 -40 -15 10 35 60 85 INPUT VOLTAGE (V) TEMPERATURE (°C) SUPPLY CURRENT **REGION OF STABLE COUT ESR** vs. TEMPERATURE vs. LOAD CURRENT 10 100 $I_{LOAD} = 50 \text{mA}$ 90 80 SUPPLY CURRENT (µA) 70 1 $C_{OUT} ESR (\Omega)$ 60 50 40 STABLE REGION 0.1 30 20 10 0 0.01 -40 -20 0 20 40 60 80 100 20 40 60 120 0 80 100 TEMPERATURE (°C) LOAD CURRENT (mA)

**MAX1749** 

M/IXI/N



M \X \M

## **Pin Description**

PIN	NAME	FUNCTION
1	ON	Active-High On/Off Input. Apply a logic high to deliver power to the load. Apply a logic low to disconnect the load and reduce the supply current to 0.1nA.
2	GND	Ground. This pin also functions as a heatsink. Solder to large pads or the circuit board ground plane to max- imize thermal dissipation.
3	IN	Regulator Input. Supply voltage can range from +2.5V to +6.5V. Bypass with $1\mu$ F to GND (see <i>Capacitor Selection and Regulator Stability</i> ).
4	OUT	Regulator Output. Fixed 1.25V or adjustable from 1.25V to V <sub>IN</sub> . Sources up to 120mA. Bypass with a 1µF, <0.2 $\Omega$ typical ESR capacitor to GND.
5	SET	Feedback Input for Setting the Output Voltage. Connect to OUT for 1.25V regulated output (see <i>Output Voltage Selection</i> ). Connect to an external resistor-divider for adjustable-output operation.

### **Detailed Description**

The MAX1749 is a low-quiescent-current, vibrator motor driver designed for battery-powered wireless handsets and pagers. The device supplies an adjustable +1.25V to +6.5V output for load currents up to 120mA. The MAX1749 allows for a constant vibration force while operating from a +2.5V to +6.5V input voltage range.

The 1.25V bandgap reference is connected to the error amplifier's inverting input. The error amplifier compares this reference with the feedback voltage and amplifies the difference. The MOSFET driver reads the error signal and applies the appropriate drive to the p-channel pass transistor. If the feedback voltage is lower than the reference voltage, the pass-transistor gate is pulled lower than the reference, allowing more current to flow and increasing the output voltage. If the feedback voltage is too high, the pass-transistor gate is pulled-up, allowing less current to flow to the output. The output voltage is fed back to SET either directly for a 1.25V fixed output or through an external resistor-divider for an adjustable +1.25V to V<sub>IN</sub> output. Additional blocks include a current limiter, reverse battery protection, a thermal sensor, and ON/ $\overline{OFF}$  logic.

#### **Internal P-Channel Pass Transistor**

The MAX1749 features a 1.1 $\Omega$  typical P-channel MOS-FET pass transistor. This provides several advantages over similar designs using PNP pass transistors, including longer battery life.



Figure 1. Functional Diagram



Figure 2. Adjustable Output Using External Feedback Resistors

The P-channel MOSFET requires no base drive current, which reduces quiescent current considerably. PNPbased regulators waste considerable amounts of base current under large loads. The MAX1749 does not suffer from these problems and consumes only 80µA of quiescent current independent of the load (see *Typical Operating Characteristics*).

#### **Output Voltage Selection**

To select the preset 1.25V output, connect OUT directly to SET. To adjust the output (1.25V to 6.5V), use two external resistors connected as a voltage divider to SET (Figure 2). The output voltage is set by the following equation:

$$V_{OUT} = V_{SET} (1 + R1 / R2)$$

where  $V_{SET} = 1.25V$ . To simplify resistor selection:

$$R1 = R2 \left( \frac{V_{OUT}}{V_{SET}} - 1 \right)$$

Choose R2 =  $100k\Omega$  to optimize power consumption, accuracy, and high-frequency power-supply rejection. The total current through the external resistive feedback and load should not be less than  $10\mu$ A.

#### ON/OFF

Drive ON high to provide power to the load. Drive ON low to disable power to the load and reduce the supply current to typically 0.1nA (1µA max). Refer to the ON/OFF waveforms in the *Typical Operating Characteristics*. When ON goes high, output current rises to the current limit until V<sub>OUT</sub> reaches regulation. While in regulation, the output current drops to a lower value sufficient to maintain motor speed. When ON goes low,

the regulator turns off; however, inertial energy in the motor exhibits a slow output voltage decline. The MAX1749 is designed to withstand this condition with no negative effects.

#### **Current Limit**

The MAX1749 includes a current limiter that monitors and controls the pass transistor's gate voltage, estimating the output current and limiting it to about 280mA. For design purposes, the current limit should be considered 120mA (min) to 420mA (max). The output can be shorted to ground for an infinite time period without damaging the part.

#### **Thermal-Overload Protection**

Thermal-overload protection limits total power dissipation in the MAX1749. When the junction temperature exceeds  $T_J = \pm 170^{\circ}$ C, the thermal sensor sends a signal to the ON/OFF logic, turning off the pass transistor and allowing the IC to cool. The thermal sensor will turn the pass transistor on again after the IC's junction temperature cools by typically 20°C, resulting in a pulsed output during continuous thermal-overload conditions.

#### **Operating Region and Power Dissipation**

Maximum power dissipation of the MAX1749 depends on the thermal resistance of the case and circuit board, the temperature difference between the die junction and ambient air, and the rate of air flow. The power dissipation across the device is  $P = I_{OUT}$  (VIN - V<sub>OUT</sub>). The resulting maximum power dissipation is:

$$PMAX = (T_J - T_A) / \theta_{JA}$$

where (T<sub>J</sub> - T<sub>A</sub>) is the temperature difference between the MAX1749 die junction and the surrounding air, and  $\theta_{JA}$  is +140°C/W.

GND performs the dual function of providing an electrical connection to ground and channeling heat away. Connect GND to a large pad or ground plane.

#### **Reverse Battery Protection**

The MAX1749 has a unique protection scheme that limits the reverse supply current to less than 1mA when either  $V_{IN}$  or  $V_{ON}$  falls below ground. The circuitry monitors the polarity of these two pins, disconnecting the internal circuitry and parasitic diodes when the battery is reversed. This feature prevents the device from overheating and damaging the battery.

#### VIN > 5.5V Minimum Load Current

When operating the MAX1749 with an input voltage above 5.5V, the minimum current through the external feedback resistors and load must be 30µA.



MAX1749

**Chip Information** 

TRANSISTOR COUNT: 148

### Capacitor Selection and Regulator Stability

Use a 1µF capacitor on the input and a 1µF capacitor on the output of the MAX1749. A higher-value input capacitor (10µF) may be necessary if large, fast transients are anticipated and the device is located several inches from the power source. Improve load-transient response and stability by using larger output capacitors. For stable operation over the full temperature range with load currents up to 120mA, use a 1µF min capacitor (ESR < 0.2 $\Omega$ ).

### Package Information



NOTES

Maxim cannot assume responsibility for use of any circuitry other than circuitry entirely embodied in a Maxim product. No circuit patent licenses are implied. Maxim reserves the right to change the circuitry and specifications without notice at any time.

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