

Data Sheet

10 GHz to 20 GHz, GaAs, MMIC, Double Balanced Mixer

HMC554ALC3B

FEATURES

Conversion loss: 8.5 dB LO to RF isolation: 37 dB Input IP3: 20 dBm RoHS compliant, 2.90 mm × 2.90 mm, 12-terminal LCC package

APPLICATIONS

Microwave and very small aperture terminal (VSAT) radios Test equipment

Military electronic warfare (EW); electronic countermeasure (ECM); and command, control, communications and intelligence (C3I)

GENERAL DESCRIPTION

The HMC554ALC3B is a general-purpose, double balanced mixer in a leadless RoHS compliant leadless chip carrier (LCC) package that can be used as an upconverter or downconverter between 10 GHz and 20 GHz. This mixer is fabricated in a gallium arsenide (GaAs) metal semiconductor field effect transistor (MESFET) process and requires no external

FUNCTIONAL BLOCK DIAGRAM



components or matching circuitry. The HMC554ALC3B provides excellent local oscillator (LO) to RF and LO to intermediate frequency (IF) isolation due to optimized balun structures. The RoHS compliant HMC554ALC3B eliminates the need for wire bonding and is compatible with high volume surface-mount manufacturing techniques.

Rev. A

Document Feedback

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TABLE OF CONTENTS

Features 1
Applications1
Functional Block Diagram1
General Description
Revision History
Specifications
Absolute Maximum Ratings 4
Thermal Resistance
ESD Caution
Pin Configuration and Function Descriptions
Interface Schematics
Typical Performance Characteristics
Downconverter Performance, IF = 100 MHz6

REVISION HISTORY

10/2019	-Rev. 0 to Rev	A
~1		

Changes to 10 GHz to 20 GHz Performance, Downconverter,
Input 1 dB Compression Point Parameter, Table 1 and 12 GHz to
16 GHz Performance, Downconverter, Input 1 dB Compression
Point Parameter, Table 1 3
Changes to Figure 13 and Figure 157
Changes to Figure 27 and Figure 30 10

4/2018—Revision 0: Initial Version

Downconverter Performance, IF = 3000 MHz	10
Upconverter Performance, $IF_{IN} = 100 \text{ MHz}$	13
Upconverter Performance, IF_{IN} = 3000 MHz	16
Isolation and Return Loss	19
IF Bandwidth—Downconverter	
Spurious and Harmonics Performance	
Theory of Operation	
Applications Information	
Typical Application Circuit	
Evaluation PCB Information	
Outline Dimensions	
Ordering Guide	

SPECIFICATIONS

 $T_A = 25^{\circ}C$, IF = 100 MHz, LO = 13 dBm, upper side band. All measurements performed as a downconverter, unless otherwise noted, on the evaluation printed circuit board (PCB).

Parameter	Symbol	Min	Тур	Max	Unit
FREQUENCY					
RF Pin		10		20	GHz
IF Pin		DC		6	GHz
LO Pin		10		20	GHz
LO AMPLITUDE		9	13	15	dBm
10 GHz TO 20 GHz PERFORMANCE					
Downconverter					
Conversion Loss			8.5	11.5	dB
Single Sideband Noise Figure	SSB NF		9.5		dB
Input Third-Order Intercept	IP3	19	20		dBm
Input 1 dB Compression Point	P1dB		10		dBm
Input Second-Order Intercept	IP2		46		dBm
Upconverter	IF _{IN}				
Conversion Loss			7		dB
Input Third-Order Intercept	IP3		19.5		dBm
Input 1 dB Compression Point	P1dB		10		dBm
Isolation					
RF to IF		24	41		dB
LO to RF		25	37		dB
LO to IF		23	41		dB
12 GHz TO 16 GHz PERFORMANCE					
Downconverter					
Conversion Loss			8		dB
Single Sideband Noise Figure	SSB NF		9		dB
Input Third-Order Intercept	IP3	16	19.5		dBm
Input 1 dB Compression Point	P1dB		9.5		dBm
Input Second-Order Intercept	IP2		45		dBm
Upconverter	IF _{IN}				
Conversion Loss			6.5		dB
Input Third-Order Intercept	IP3		18		dBm
Input 1 dB Compression Point	P1dB		10		dBm

ABSOLUTE MAXIMUM RATINGS

Table 2.

Parameter	Rating
RF Input Power	25 dBm
LO Input Power	26 dBm
IF Input Power	25 dBm
IF Source/Sink Current	3 mA
Reflow Temperature	260°C
Maximum Junction Temperature	175°C
Continuous Power Dissipation, P_{DISS} ($T_A = 85^{\circ}$ C, Derate 3.7 mW/°C Above 85°C)	333 mW
Operating Temperature Range	-40°C to +85°C
Storage Temperature Range	–65°C to +150°C
Electrostatic Discharge (ESD) Sensitivity	
Human Body Model (HBM)	250 V; Class 0B
Field Induced Charged Device Model (FICDM)	1250 V; Class IV

Stresses at or above those listed under Absolute Maximum Ratings may cause permanent damage to the product. This is a stress rating only; functional operation of the product at these or any other conditions above those indicated in the operational section of this specification is not implied. Operation beyond the maximum operating conditions for extended periods may affect product reliability.

THERMAL RESISTANCE

Thermal performance is directly linked to PCB design and operating environment. Careful attention to PCB thermal design is required.

 θ_{JA} is the natural convection junction to ambient thermal resistance measured in a one cubic foot sealed enclosure. θ_{JC} is the junction to case thermal resistance.

Table 3. Thermal Resistance

Package Type	θ」Α	ονθ	Unit		
E-12-4 ¹	120	195	°C/W		

¹ Test Condition 1: JEDEC standard JESD51-2.

ESD CAUTION



ESD (electrostatic discharge) sensitive device. Charged devices and circuit boards can discharge without detection. Although this product features patented or proprietary protection circuitry, damage may occur on devices subjected to high energy ESD. Therefore, proper ESD precautions should be taken to avoid performance degradation or loss of functionality.

PIN CONFIGURATION AND FUNCTION DESCRIPTIONS



Figure 2. Pin Configuration

Table 4. Pin Function Descriptions

Pin No.	Mnemonic	Description
1, 3, 4, 6, 7, 9	GND	Ground. These pins and package bottom must be connected to RF/dc ground.
2	LO	LO Port. This pin is ac-coupled and matched to 50 Ω .
5	IF	IF Port. This pin is dc-coupled. For applications not requiring operation to dc, dc block this port externally using a series capacitor of a value chosen to pass the necessary IF frequency range. For operation to dc, this pin must not source/sink more than 3 mA of current or die malfunction and possible die failure may result.
8	RF	RF Port. This pin is ac-coupled and matched to 50 Ω .
10, 11, 12	NIC	Not Internally Connected. These pins can be connected to RF/dc ground. Performance is not affected.
	EPAD	Exposed Pad. The exposed pad must be connected to RF/dc ground.

INTERFACE SCHEMATICS



Figure 3. GND Interface Schematic



Figure 4. LO Interface Schematic



Figure 5. IF Interface Schematic



Figure 6. RF Interface Schematic

TYPICAL PERFORMANCE CHARACTERISTICS

DOWNCONVERTER PERFORMANCE, IF = 100 MHz

















Figure 10. Conversion Gain vs. RF Frequency at Various LO Power Levels, $T_A = 25^{\circ}$ C



Figure 11. Input IP3 vs. RF Frequency at Various LO Power Levels, $T_A = 25^{\circ}C$



Figure 12. Noise Figure vs. RF Frequency at Various LO Power Levels, $T_A = 25^{\circ}C$

Data Sheet

INPUT P1dB (dBm) $\begin{array}{l} \mathsf{T}_\mathsf{A} = +85^\circ\mathsf{C} \\ \mathsf{T}_\mathsf{A} = +25^\circ\mathsf{C} \\ \mathsf{T}_\mathsf{A} = -40^\circ\mathsf{C} \end{array}$ 13895-015 RF FREQUENCY

Figure 13. Input P1dB vs. RF Frequency at Various Temperatures, LO = 13 dBm



Figure 14. Input IP2 vs. RF Frequency at Various Temperatures, LO = 13 dBm



Figure 15. Input P1dB vs. RF Frequency at Various LO Power Levels, $T_A = 25^{\circ}$ C



Figure 16. Input IP2 vs. RF Frequency at Various LO Power Levels, $T_A = 25^{\circ}C$

4

2

0

10 11 12 13 14 15 16 17 18 19 20

T_A = +85°C

T_A = +25°C

 $T_A = -40^{\circ}C$

Lower Sideband (High-Side LO)



Figure 17. Conversion Gain vs. RF Frequency at Various Temperatures, LO = 13 dBm



RF FREQUENCY

Figure 19. Noise Figure vs. RF Frequency at Various Temperatures, LO = 13 dBm



Figure 20. Conversion Gain vs. RF Frequency at Various LO Power Levels, $T_A = 25^{\circ}$ C



Figure 21. Input IP3 vs. RF Frequency at Various LO Power Levels, $T_A = 25^{\circ}C$



Figure 22. Noise Figure vs. RF Frequency at Various LO Power Levels, $T_A = 25^{\circ}$ C

3895-019

Data Sheet





Figure 24. Input IP2 vs. RF Frequency at Various LO Power Levels, $T_A = 25^{\circ}$ C

DOWNCONVERTER PERFORMANCE, IF = 3000 MHz





Figure 25. Conversion Gain vs. RF Frequency at Various Temperatures, $LO = 13 \, dBm$





CONVERSION GAIN (dB)



Figure 29. Input IP3 vs. RF Frequency at Various LO Power Levels, $T_A = 25^{\circ}C$



Figure 30. Input P1dB vs. RF Frequency at Various LO Power Levels, $T_A = 25^{\circ}C$



RF FREQUENCY

16 17 18 19 20

T_A = +25°C $T_A = -40^{\circ}C$

12

13 14 15

0

10 11 3895-029

Data Sheet







Figure 32. Input IP2 vs. RF Frequency at Various LO Power Levels, $T_A = 25^{\circ}$ C

Lower Sideband (High-Side LO)



Figure 33. Conversion Gain vs. RF Frequency at Various Temperatures, LO = 13 dBm





Figure 36. Conversion Gain vs. RF Frequency at Various LO Power Levels, $T_A = 25^{\circ}$ C



Figure 37. Input IP3 vs. RF Frequency at Various LO Power Levels, $T_A = 25^{\circ}C$



Figure 38. Input IP2 vs. RF Frequency at Various LO Power Levels, $T_A = 25^{\circ}$ C

UPCONVERTER PERFORMANCE, $IF_{IN} = 100 \text{ MHz}$



Figure 39. Conversion Gain vs. RF Frequency at Various Temperatures, LO = 13 dBm



Figure 40. Input IP3 vs. RF Frequency at Various Temperatures, LO = 13 dBm



Figure 41. Input P1dB vs. RF Frequency at Various Temperatures, LO = 13 dBm



Figure 42. Conversion Gain vs. RF Frequency at Various LO Power Levels, $T_A = 25^{\circ}$ C



Figure 43. Input IP3 vs. RF Frequency at Various LO Power Levels, $T_A = 25^{\circ}C$



Figure 44. Input P1dB vs. RF Frequency at Various LO Power Levels, $T_A = 25^{\circ}$ C







Figure 46. Input IP2 vs. RF Frequency at Various LO Power Levels, $T_A = 25^{\circ}$ C

Lower Sideband (High-Side LO)



Figure 47. Conversion Gain vs. RF Frequency at Various Temperatures, LO = 13 dBm



Figure 48. Input IP3 vs. RF Frequency at Various Temperatures, LO = 13 dBm







Figure 50. Conversion Gain vs. RF Frequency at Various LO Power Levels, $T_{\rm A} = 25\,^{\circ}{\rm C}$



Figure 51. Input IP3 vs. RF Frequency at Various LO Power Levels, $T_A = 25^{\circ}C$



Figure 52. Input IP2 vs. RF Frequency at Various LO Power Levels, $T_A = 25^{\circ}$ C

UPCONVERTER PERFORMANCE, IF_{IN} = 3000 MHz





Figure 53. Conversion Gain vs. RF Frequency at Various Temperatures, $LO = 13 \, dBm$





Figure 55. Input P1dB vs. RF Frequency at Various Temperatures, LO = 13 dBm



Figure 56. Conversion Gain vs. RF Frequency at Various LO Power Levels, $T_A = 25^{\circ}$ C



Figure 57. Input IP3 vs. RF Frequency at Various LO Power Levels, $T_A = 25^{\circ}$ C



Figure 58. Input P1dB vs. RF Frequency at Various LO Power Levels, $T_A = 25^{\circ}$ C

Data Sheet







Figure 60. Input IP2 vs. RF Frequency at Various LO Power Levels, $T_A = 25^{\circ}$ C

Lower Sideband (High-Side LO)



Figure 61. Conversion Gain vs. RF Frequency at Various Temperatures, LO = 13 dBm



Figure 62. Input IP3 vs. RF Frequency at Various Temperatures, LO = 13 dBm



Figure 63. Input IP2 vs. RF Frequency at Various Temperatures, LO = 13 dBm



Figure 64. Conversion Gain vs. RF Frequency at Various LO Power Levels, $T_A = 25^{\circ}$ C



Figure 65. Input IP3 vs. RF Frequency at Various LO Power Levels, $T_A = 25^{\circ}C$



Figure 66. Input IP2 vs. RF Frequency at Various LO Power Levels, $T_A = 25^{\circ}$ C

ISOLATION AND RETURN LOSS



Figure 67. LO to RF Isolation vs. RF Frequency at Various Temperatures, $LO = 13 \, dBm$



Figure 68. LO to IF Isolation vs. RF Frequency at Various Temperatures, $LO = 13 \ dBm$



Figure 69. RF to IF Isolation vs. RF Frequency at Various Temperatures, LO = 13 dBm



Figure 70. LO to RF Isolation vs. RF Frequency at Various LO Power levels, $T_{\rm A} = 25^{\circ}{\rm C}$



Figure 71. LO to IF Isolation vs. RF Frequency at Various LO Power Levels, $T_A = 25^{\circ}$ C



Figure 72. RF to IF Isolation vs. RF Frequency at Various LO Power Levels, $T_A = 25^{\circ}$ C



Figure 74. RF Return Loss vs. RF Frequency at LO Power Levels, $T_A = 25^{\circ}$ C, LO = 15 GHz



Figure 75. IF Return Loss vs. IF Frequency at LO Power Levels, $T_A = 25$ °C, LO = 15 GHz

IF BANDWIDTH—DOWNCONVERTER





Figure 76. Conversion Gain vs. IF Frequency at Various Temperatures, LO = 13 dBm



Figure 77. Input IP3 vs. IF Frequency at Various Temperatures, $LO = 13 \ dBm$



Figure 78. Conversion Gain vs. IF Frequency at Various LO Power Levels, $T_A = 25^{\circ}C$



Figure 79. Input IP3 vs. IF Frequency at Various LO Power Levels, $T_{\rm A} = 25\,^{\circ}{\rm C}$

Lower Sideband, LO Frequency = 19 GHz



Figure 80. Conversion Gain vs. IF Frequency at Various Temperatures, $LO = 13 \, dBm$





Figure 82. Conversion Gain vs. IF Frequency at Various LO Power Levels, $T_A = 25^{\circ}$ C



Figure 83. Input IP3 vs. IF Frequency at Various LO Power Levels, $T_A = 25^{\circ}C$

SPURIOUS AND HARMONICS PERFORMANCE

Mixer spurious products are measured in dBc from the IF output power level. N/A means not applicable.

LO Harmonics

LO = 13 dBm, all values in dBc below input LO level and measured at RF port.

Table 5. LO Harmonics at RF

	N × LO Spur at RF Port			
LO Frequency (GHz)	1	2	3	4
12	39	39	59	57
13	38	40	70	N/A
15	38	48	49	N/A
16	37	56	50	N/A
18	36	54	N/A	N/A
19	36	53	N/A	N/A
21	36	46	N/A	N/A

LO = 13 dBm, all values in dBc below input LO level and measured at IF port.

Table 6. LO Harmonics at IF

		N × LO Spur at IF Port			
LO Frequency (GHz)	1	2	3	4	
12	38	77	67	89	
13	41	63	74	N/A	
15	44	72	56	N/A	
16	42	53	56	N/A	
18	44	79	N/A	N/A	
19	53	70	N/A	N/A	
21	47	75	N/A	N/A	

M × **N** Spurious Outputs

Downconverter, Upper Sideband

Spur values are $(M \times RF) - (N \times LO)$.

RF = 15.1 GHz at -10 dBm, LO = 15 GHz at 13 dBm.

		N × LO					
		0 1 2 3 4 5					
	0	N/A	14	47	27	N/A	N/A
M×RF	1	48	0	70	72	65	N/A
	2	75	77	60	79	74	68
	3	65	74	79	70	78	71
	4	N/A	60	74	80	88	78
	5	N/A	N/A	56	72	81	88

Upconverter, Upper Sideband

Spur values are $(M \times IF) + (N \times LO)$.

 $IF_{IN} = 100 MHz at -10 dBm$, LO = 15 GHz at 13 dBm.

		N × LO			
		0	1	2	3
	-5	89	80	73	67
	-4	88	79	73	68
	-3	91	66	74	66
	-2	91	67	74	66
	-1	36	0	35	20
M×IF	0	N/A	6	17	22
	+1	36	0	35	19
	+2	88	63	73	65
	+3	90	63	74	66
	+4	90	80	73	65
	+5	88	78	72	66

THEORY OF OPERATION

The HMC554ALC3B is a general-purpose, double balanced mixer that can be used as an upconverter or a downconverter from 10 GHz to 20 GHZ.

When used as a downconverter, the HMC554ALC3B downconverts RF between 10 GHz and 20 GHz to IF between dc and 6 GHz.

When used as an upconverter, the mixer upconverts intermediate frequencies between dc and 6 GHz to radio frequencies between 10 GHz and 20 GHz.

APPLICATIONS INFORMATION TYPICAL APPLICATION CIRCUIT

Figure 84 shows the typical application circuit for the HMC554ALC3B. The HMC554ALC3B is a passive device and does not require any external components. The IF pin is internally dc-coupled. The RF and LO pins are internally ac-coupled. When IF operation to dc is not required, using an external series capacitor is recommended, of a value chosen to pass the necessary IF frequency range. When IF operation to dc is required, do not exceed the IF source and sink current rating specified in the Absolute Maximum Ratings section.



Figure 84. Typical Application Circuit

EVALUATION PCB INFORMATION

Use RF circuit design techniques for the circuit board used in the application. Ensure that signal lines have 50 Ω impedance and connect the package ground leads and the exposed pad directly to the ground plane (see Figure 84). Use a sufficient number of via holes to connect the top and bottom ground planes. The evaluation circuit board shown in Figure 85 is available from Analog Devices, Inc., upon request.

Table 7. List of Materials for Evaluation PCB
EV1HMC554ALC3B

2,111,1000,112,002				
ltem	Description			
J1, J2	PCB mount SRI 2.92 mm connectors			
J3	PCB mount Johnson SMA connector			
U1	HMC554ALC3B			
PCB ¹	117611-1 evaluation board on Rogers 4350			

¹ 117611-1 is the raw bare PCB identifier. Reference EV1HMC554ALC3B when ordering complete evaluation PCB.



Figure 85. Evaluation PCB Top Layer

OUTLINE DIMENSIONS



Figure 86. 12-Terminal Ceramic Leadless Chip Carrier (LCC, (E-12-4) Dimensions shown in millimeters

ORDERING GUIDE

Model ¹	Temperature Range	MSL Rating ²	Package Description	Package Option
HMC554ALC3B	–40°C to +85°C	MSL3	12-Terminal Ceramic [LCC]	E-12-4
HMC554ALC3BTR	–40°C to +85°C	MSL3	12-Terminal Ceramic [LCC]	E-12-4
HMC554ALC3BTR-R5	–40°C to +85°C	MSL3	12-Terminal Ceramic [LCC]	E-12-4
EV1HMC554ALC3B			Evaluation PCB Assembly	

¹ All models are RoHS compliant.

² The peak reflow temperature is 260°C. See the Absolute Maximum Ratings section, Table 2.



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